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Powering internet-of-things from ambient energy: a review

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Abstract

Internet-of-thing (IoT) is an assembly of devices that collect and share data with other devices and communicate via the internet. This massive network of devices, generates and communicates data and is the key to the value in IoT, allowing access to raw information, gaining insight, and making an intelligent decisions. Today, there are billions of IoT devices such as sensors and actuators deployed. Many of these applications are easy to connect, but those tucked away in hard-to-access spots will need to harvest ambient energy. Therefore, the aim is to create devices that are self-report in real-time. Efforts are underway to install a self-powered unit in IoT devices that can generate sufficient power from environmental conditions such as *light*, *vibration*, and *heat*. In this review paper, we discuss the recent progress made in materials and device development in power- and storage units, and power management relevant for IoT applications. This review paper will give a comprehensive overview for new researchers entering the field of IoT and a collection of challenges as well as perspectives for people already working in this field.

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1. Introduction

The internet of things (IoT) is a giant network that connects a huge number of physical objects—‘things’—through the internet. Remote accessibility and automatization allow us to perform tasks efficiently and repeatedly. IoT is considered one of the leading technologies, which is projected to increase and reach 50 billion devices by the year 2030 [1, 2]. Applications of IoT are primarily based on low-power consumable devices, which are used in home appliances or automation [3], hospitals [1], and healthcare [4, 5] as well as in numerous industrial processes [6]. Typically, IoT devices receive or generate data and then transmit it wirelessly to other devices to perform any given task. It consists of mainly three parts: IoT units (wireless sensor networks (WSNs), processors, actuators, etc), power units, and electronics.

Powering billions of devices remains an open technological challenge. At the same time, the installation of billions of devices will require the miniaturization of IoT units, power units, and electronics. Therefore, IoT-based industries face several technological challenges: (a) creating low-power electronics, including wireless communications, sensors, and actuators, (b) producing reliable and maintenance-free autonomous powering devices with a long life span, (c) developing low-cost fabrication technology that supports the miniaturization of billions of IoT units, and (d) creating reliable and safe wireless communication methods [7–10].

The current technological solution for powering IoT-based devices mainly relies on the battery industries [11–15]. However, their lifetimes are generally much less than the expected lifetimes of the WSNs and hence, replacement of batteries is required periodically to keep the devices operational. This creates extra expenses and additional complications for remote sensors, and in some cases, it can even be impossible to replace batteries. Moreover, batteries are expensive, bulky, and contain harmful chemicals. Although an effort is currently taken to improve the energy storage capacity and therefore the lifetime of IoT devices, the miniaturization of batteries remains a major technological challenge.

As an alternative, a self-powered unit can be installed in the IoT devices, which can harness sufficient power from ambient energy sources [16]. Usually, a few energy conversion technologies are available to date which can convert various sources of ambient energy into electricity. Some examples are a) thermoelectric (TE) or pyroelectric generators, which can harness energy from ambient waste heat, b) piezoelectric, flexoelectric, and electromagnetic vibration energy devices which can convert mechanical (vibrational) energy to electric voltage; c) photovoltaic devices, which can convert light into electricity etc. Depending on the available ambient energy sources and conversion efficiencies, these energy-harvesting technologies can power varieties of IoT devices, as shown in figure 1. The energy demand for low-power IoT devices typically lies between several microwatts (μWs) to a few milliwatts (mWs) or between $10\text{--}1000\ \mu\text{W cm}^{-2}$ in terms of area-specific power density [17].

In this review paper, we report the current developments in the above mentioned important powering- and storage units, and the electronics of IoT devices. The paper is divided into three sections: section 2 describes energy harvesting devices, which include energy harvesting only from ambient heat, vibration, and light. Section 3 describes energy storage devices, which is mainly devoted to batteries. The use of micro-fuel cells for on-demand powering of IoT devices is included in this section. Section 4 describes the electronic power management for IoT devices.

It is noteworthy that other energy harvester- (wind- [17, 18], bio-waste- [19, 20], electromagnetic radiation energy [21]), storage devices (supercapacitors [22–28]) and the economic competitiveness of these devices for IoT applications [29–31] are not discussed here due to the large volume of this article. Above referred articles might help readers to dive deeper into each topic.

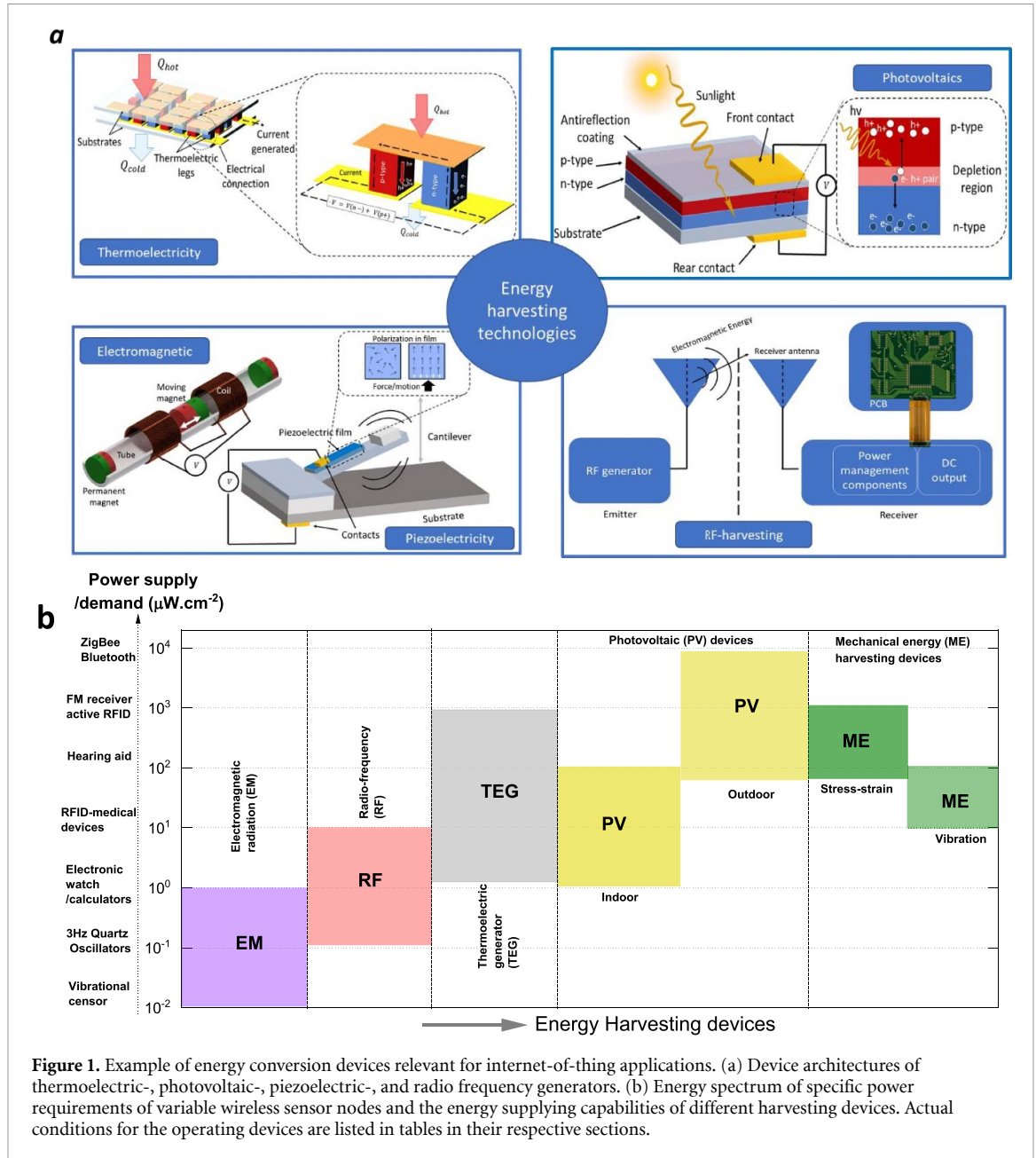
2. Section-I: energy harvesting devices

In this section, we present five different types of harvesting devices, which include thermoelectricity, piezoelectricity, flexo-electricity, electromagnetic vibration energy harvesters, and photovoltaics (PV).

2.1. Thermoelectricity for IoT applications

A few comprehensive review articles are published in the literature based on the μ -TE generators [32, 33], wearable TE generators [34–36], and the use of TE devices for the IoT applications [29, 37]. Here, we have updated the current developments on the materials and micro-devices, which are potential for IoT applications.

TE devices are considered as one of the attractive solutions for powering IoT-based WSNs due to their long operational lifetime, high reliability, and more importantly, their maintenance-free characteristics [38]. These devices are capable of converting electric voltage output directly from ambient heat sources using the Seebeck effect, without involving any mechanical part on it. Typically, a thermoelectric generator (TEG)



consists of both p- and n-type semiconductors (that is, p–n thermocouples) that are connected electrically in a series but thermally in parallel [39]. The maximum achievable conversion efficiency of such TEG (η_{TEG}) is given by equation (1) [40]:

$$\eta_{\text{TEG}} = \frac{T_{\text{Hot}} - T_{\text{Cold}}}{T_{\text{Hot}}} \frac{\sqrt{1 + ZT} - 1}{\sqrt{1 + ZT} + \frac{T_{\text{Cold}}}{T_{\text{Hot}}}}, \quad (1)$$

The overall efficiency depends on (a) the Carnot efficiency ($\frac{T_{\text{Hot}} - T_{\text{Cold}}}{T_{\text{Hot}}}$), and (b) the TE figure of merit (ZT). The quality of a TE material is determined by the factor ZT ($ZT = \frac{\sigma S^2}{\kappa} T$; where σ , S , κ , and T are electrical conductivity, Seebeck coefficient, thermal conductivity, and the mean temperature between the hot, and the cold end, respectively). Therefore, it suggests that an appropriate design of the device for the propagation of heat flux from the hot to the cold end [41], and the use of high ZT materials are essential to achieve high-energy conversion efficiency. However, the output power is equally important as the efficiency for some miniature devices (such as IoT-based WSNs, micro-temperature sensors, etc.) [42] because the amount of heat flow from the hot to the cold reservoir in such devices is very small [43]. Depending on the ambient temperature sources and the materials used, TEGs are capable of supplying power to those WSNs in a spectrum that lies between 10 and 1000 $\mu\text{W cm}^{-2}$ as shown in figure 1(b) (see also figures 18–20 in [37]).

2.1.1. Materials

Implementation of both p- and n-type semiconducting materials with high ZT is necessary to achieve maximum conversion efficiency in a TEG. Obtaining materials with high ZT is challenging because (a) the electronic conductivity and the electronic part of the thermal conductivities are interrelated to each other through the density of charge carriers according to Wiedemann–Franz law [44], and (b) in a band-gap material, the electrical conductivity and the Seebeck coefficients behave oppositely as a function of the density of charge carriers [45]. Hence, optimization of the density of charge carriers and the minimization of the lattice part of the thermal conductivity is required, independently [44, 46]. As shown in figure 2, the library of TE materials shows many families of compounds with $0.5 \leq ZT < 3$ [47]. Examples include Skutterudites [48], half-Heusler [49, 50] clathrates [51], zintl-phase [52–54], silicates [55], silicon-germanium [56], heavy metal-based chalcogenides [57, 58], transition-metal-based oxides [59–61] etc. Recently, polycrystalline SnSe-based chalcogenides exhibit ZT slightly above 3 (i.e. $2 < ZT_{\max} < 3.2$) [62–64] but only at higher temperatures (~ 800 – 1000 K). Therefore, their use for near room-temperature applications remains unattainable. Additionally, not all of these materials have been implemented in devices and therefore, additional challenges associated with electrical contact resistance [65] and thermal contact resistances [66] are not known. Recently, half-Heusler-based thin films grown on silicon substrates show an estimated $ZT_{\max} \approx 6$ near 75°C [67]). This value is extremely high due to the fact that the figure of merit was calculated by combining power factor measured in the in-plane direction together with the thermal conductivity measured in the out-of-plane direction. Nonetheless, although this value should be taken with caution, it is interesting to point out that films were deposited on silicon substrates. n-type organic TE materials also exhibit potential figure-of-merit (FOM) ($ZT > 0.3$ at 120°C [68]), which can be used to convert energy from waste heat of a human body (i.e. in wearable devices).

2.1.2. Devices

Appropriate design of the device architecture allows for improving the device performance by reducing electrical- and thermal contact resistance [69]. On the other hand, technologies that allow a high-density integration of thermocouples can generate sufficient power output even by using a low ZT material (such as silicon) [42]. As the power output from a TEG device is more relevant for IoT-based applications [42], our discussion includes the progress on the overall power output but not the overall efficiency. The power output of a TEG is proportional to the cross-section area of the device (A) and the square of the applied temperature gradient ($(\Delta T)^2$). Usually, two parameters are used in the literature to quantify the power output capacity of TEGs; area-specific power density (i.e. $P_A = P_{\max}/A$) and specific power generation capacity ($\Gamma_P = P_{\max}/[A(\Delta T)^2]$).

A list of commercial TEGs and their performance can be found in [29, 37]. The power output of some commercial TEGs can be as high as 0.69 W cm^{-2} , where the temperature at the hot end and the footprint area were 500°C and $5.6 \times 5.6\text{ cm}^2$, respectively [29]. This indicates that commercial TEGs can power a long-range of IoT-based sensors [37]. However, the footprint area of IoT-based devices shrinks continuously ($< 1\text{ cm}^3$) [37, 70] and so, more attention is paid to the miniaturization of TEGs [43, 69]. In this sense, the development of cost-effective micro-devices (μ -TEGs) is becoming significantly important.

Depending on the propagation of heat from the hot to the cold reservoir, μ -TEGs devices can have three different architectures; planar, vertical, and hybrid [32, 33]. Thermocouples are suspended on a membrane (or substrates) in a planar TEG device and heat propagates parallel to the substrates. The membranes used in this design often show poor mechanical stability and a large parasitic heat transfer but remain technologically important as are compatible with the silicon integrated circuit technology. Temperature gradients and the power output in this architecture can be tuned by optimizing the length and the thickness of thermocouples. In a vertical μ -TEG, thermocouples are placed vertically and so heat propagates vertically to the substrate. A large density of thermocouples can be integrated with this device layout and high power output can be achieved, which can further be tuned by controlling the width of the thermocouples [69]. Some of the common features of the vertical and planar designs are merged in a hybrid configuration where heat flows out-of-plane of the substrate but current flow in-plane and therefore have some more flexibilities [71]. In this layout, cavities are created in the substrate below the thermocouple pair to further suppress the heat flow [29, 32, 71]. The performances of a long list of μ -TEGs with different device layouts and with different TE materials can be found in [32] and some recent advancements are shown in table 1 in this article.

Incompatibility of exotic TE materials such as Bi_2Te_3 , PbTe , SnSe , etc into silicon technology prevents their applications in the area of IoT and microelectronics. Only a few silicon-based materials such as Si and SiGe are compatible with silicon-based micro-fabrication techniques. As discussed in the materials section, SiGe-based materials are one of the promising TE materials used in space missions [38], which work more efficiently at nanoscales especially due to the reduced total thermal conductivity [56]. Recently, SiGe nanowire-based μ -TEG fabricated by using micro-electro-mechanical system (MEMS) technology shows a

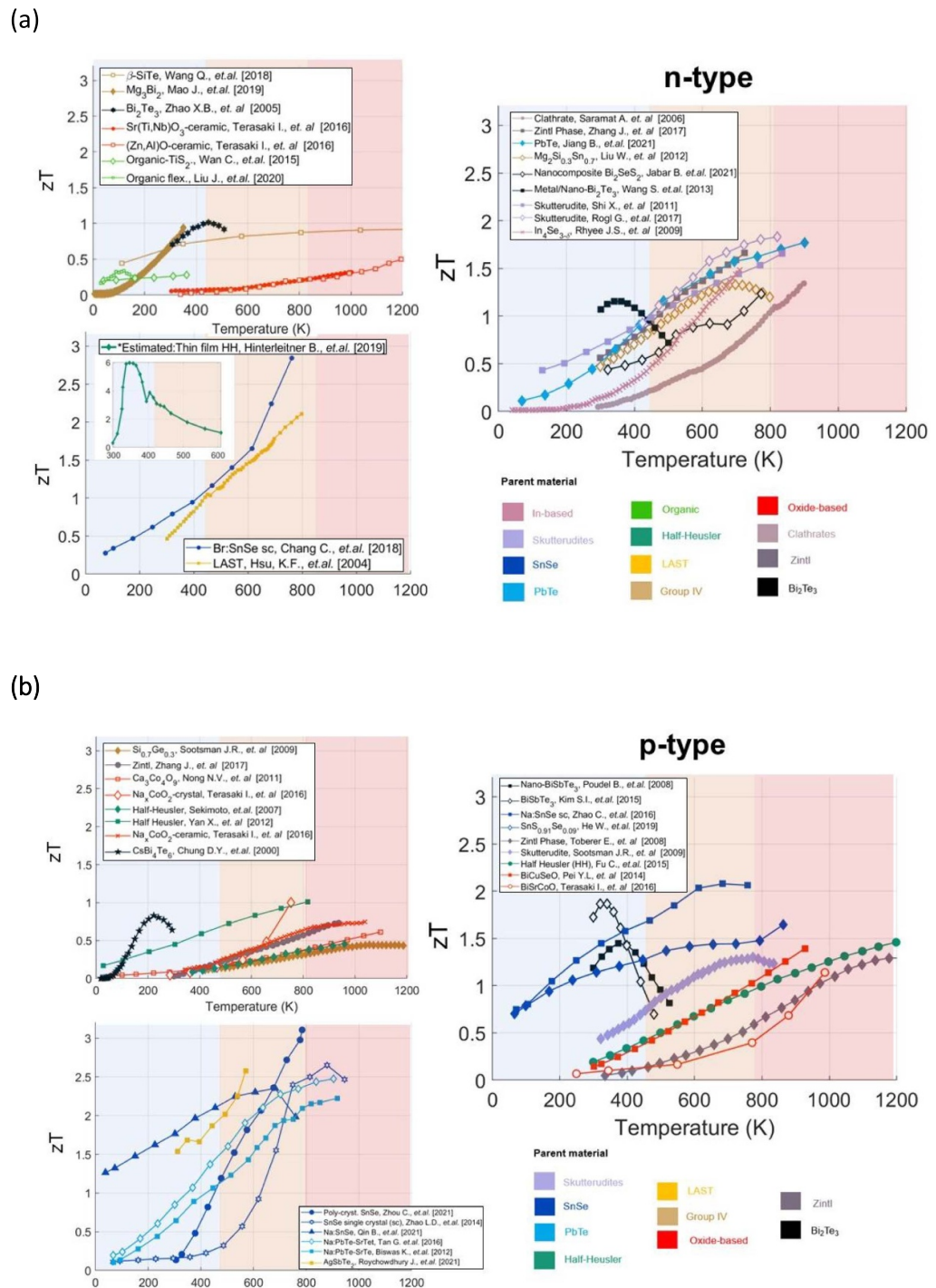


Figure 2. State-of-the-art p- and n-type thermoelectric materials with a high thermoelectric figure of merit (ZT). (a)-temperature dependence on the ZT of the best p-type materials reported to date. **Materials with $0 < ZT \leq 1$** - Zintl phase (Toberer *et al* [72]) half-Heusler (Yan *et al* [73]), half-Heusler (Sekimoto *et al* [74]), $\text{Ca}_3\text{Co}_4\text{O}_9$ ceramic (Nong *et al* [75]), Na_xCoO_2 -crystal and ceramic (Terasaki [59]), CsBi_4Te_6 (Chung *et al* [76]), $\text{p-Si}_{0.7}\text{Ge}_{0.3}$ (Sootsman *et al* [46]); **materials with $0 < ZT \leq 2$** $\text{Bi}_2\text{Sr}_2\text{Co}_2\text{O}_x$ (Terasaki [59]), BiCuSeO (Pei *et al* [77]), Nanocomposite- BiSbTe_3 (Poudel *et al* [78]), BiSbTe_3 (Kim *et al* [79]), Na:SnSe -single crystal (Zhao *et al* [80]), $\text{Sn}_{0.91}\text{Se}_{0.09}$ (Wenke *et al* [81]), Skutterudite (Sootsman *et al* [46]), half-Heusler (Fu *et al* [82]). **Materials with $0 < ZT \leq 3$** polycrystalline-SnSe (Zhou *et al* [64]), single-crystalline-SnSe (Zhao *et al* [83]), 1.5%Na-SnSe (Qin *et al* [84]), PbTe-SrTe (Tan *et al* [58]), Na-PbTe-SrTe (Biswas *et al* [85]), AgSbTe₂ (Roychowdhury *et al* [86]). (b)- temperature dependence on the ZT of the state-of-the-art n-type materials. **Materials with $0 < ZT \leq 1$** - n- Bi_2Te_3 (Zhao *et al* [87]), SiTe (Wang *et al* [88]), Mg_2Bi_2 (Mao *et al* [89]), Half-Heusler, $\text{Y}_{14}\text{Mn}_{1-x}\text{Al}_x\text{Sb}_{11}$, $\text{Sr}(\text{Ti,Nb})\text{O}_3$ and ZnO:Al (Terasaki [59]), TiS_2 (Wan *et al* [90]), Organic thermoelectric (Liu *et al* [68]); **materials with $0 < ZT \leq 2$** - Zintl phase (Zhang *et al* [91]), Clathrates (Saramat [92]), PbSe (Jiang *et al* [93]), $\text{Mg}_2\text{Si}_{0.3}\text{Sn}_{0.7}$ (Liu *et al* [94]), nanocomposite- Bi_2Se_3 (Jabar *et al* [95]), metal-nano + Bi_2Te_3 (Wang *et al* [96]), Skutterudites (Shi *et al* [97]), Skutterudites (Rogl *et al* [48]), $\text{In}_4\text{Se}_{3.8}$ crystal (Rhyee *et al* [98]); **materials with $0 < ZT \leq 3$** - half-Heusler thin film (Hinterleitner *et al* [67]), Br:SnSe single crystal (Chang *et al* [99]), LAST- $\text{AgPb}_{10}\text{SbTe}_{12}$ (Hsu *et al* [100]). Different colored regimes in the figures represent different ranges of temperatures.

Table 1. Comparison of the electrical power output and footprint areas of commercial TEGs, micro-TEGs based on the Bi₂Te₃-Sb₂Te₃ family, organic TEGs, and micro-TEGs that are compatible with MEMS and CMOS technologies. $P_A = (P_{\max}/A)$; $\Gamma_A = [P_{\max}/(A \cdot (\Delta T)^2)]$. P_{\max} and A are the maximum power output and the footprint area of the device, respectively.

Materials	Cross-section area	$P_{\max}/\Delta T$	Power density (P_A)	Γ_p ($\mu\text{W cm}^{-2} \text{K}^{-2}$)	Source
Commercial TEGs					
TGPR-22 W-7 V-56 S	$5.6 \times 5.6 \text{ cm}^2$	21.7 W	0.69 W cm^{-2}	—	[29]
HZ-2	$2.9 \times 2.9 \text{ cm}^2$	2.25 W/200 K	0.267 W cm^{-2}	6.69	[29, 106]
TGM-127-1,9-0,8	$3.0 \times 3.0 \text{ cm}^2$	5.1 W/170 K	0.57 W cm^{-2}	19.61	[29, 107]
TG12-2.5 (II-VI MARLOW)	$3.0 \times 3.0 \text{ cm}^2$	0.41 W/180 K	0.045 W cm^{-2}	1.4	[108]
Heavy metal chalcogenides: μ-TEGs					
Bi ₂ Te ₃ -Sb ₂ Te ₃	$0.5\text{--}25 \text{ cm}^2$	2.8 mW	—	83.8	[109]
Bi ₂ Te ₃ -Sb ₂ Te ₃	0.325 cm^2	2.99 mW/52.5 K	9.2 mW cm^{-2}	3.3	[110]
Bi ₂ Te ₃ /Cu, Cu annealed	1.04 cm^2	2.34 mW	2.434 mW cm^{-2}	1.63	[111]
CMOS-based fabrication technology: μ-TEGs					
Si NWs/SOG	$50 \times 50 (\mu\text{m})^2$	29.3 $\mu\text{W}/56 \text{ K}$	—	—	[112]
Si	—	0.41	—	0.48	[42]
Si NWs	$5 \times 5 \text{ mm}^2$ (TEG)	1.5 nW/0.12 K	—	—	[113]
Si NWs blades	—	−/5 K	$12 \mu\text{W cm}^{-2}$	0.48	[103]
Poly-Si(SiGe) quantum well-like str.	$60 \times 4 (\mu\text{m})^2$	−/20 K	—	0.251	[114]
Poly Si	$3 \times 3 \text{ mm}^2$	—	—	0.252	[115]
Si blades	$48 \times 36 (\mu\text{m})^2$	$0.60 \mu\text{W}/33.9 \text{ K}$	$\sim 35.0 \text{ mW cm}^{-2}$	29.0	[69]
Si(0.97)Ge(0.03)	$48 \times 36 (\mu\text{m})^2$	$0.38 \mu\text{W}/16.1 \text{ K}$	$\sim 22.0 \text{ mW cm}^{-2}$	84.0	[43]
Si(0.98)Ge(0.02)	$48 \times 36 (\mu\text{m})^2$	—	—	78.0	[43]
Si(0.99)Ge(0.01)	$48 \times 36 (\mu\text{m})^2$	—	—	52.0	[43]
Si(1.00)Ge(0.00)	$48 \times 36 (\mu\text{m})^2$	—	—	25.	[43]
MEMS-based fabrication technology: μ-TEGs					
Si NWs	2.0 mm^2 (*)	832.0 nW	$41.6 \mu\text{W cm}^{-2}$	—	[102]
SiGe NWs	2.0 mm^2	900.0 nW	$45.2 \mu\text{W cm}^{-2}$	—	[102]
Si microbeam	2.0 mm^2	690.0 nW	$34.5 \mu\text{W cm}^{-2}$	—	[102]
SiGe Nw	2.0 mm^2	142.0 nW/14.3 K	$7.1 \mu\text{W cm}^{-2}$	0.034	[116]
Silicon membrane	0.25 mm^2	11.25 nW/5.5 K	$4.5 \mu\text{W cm}^{-2}$	0.149	[101]
Polymer-based flexible TEGs					
Organic: poly[Cu _x (Cu-ett)]: p-type + poly[Na _x (Ni-ett)]: n-type	26.7 mm^2	750.0 $\mu\text{W}/82 \text{ K}$	$2.8 \mu\text{W cm}^{-2}$	4.16×10^{-4}	[117]
Organic-inorganic hybrid	0.2 mm^2	335.0 nW/20 K	1.68 W m^{-2}	4.0×10^{-4}	[118]

Note: (*) active device area.

high power density [101] ($\sim 7.1 \mu\text{W cm}^{-2}$) and was further improved (to $45.5 \mu\text{W cm}^{-2}$) by integrating a heat exchanger [102]. Similarly, power densities as high as 41.2 and $34.5 \mu\text{W cm}^{-2}$ were obtained by using μ -TEGs based on Si nanowires and Si microbeams [102], respectively (see table 1 for comparison).

Therefore, these planar devices show full technological potential for powering low-power IoT-based WSNs].

On the other hand, specific power generation capacity (Γ_p) as high as $29 \mu\text{W cm}^{-2} \text{K}^{-2}$ (estimated $P_A = 0.034 \text{ W cm}^{-2}$) has been achieved recently at room temperature in nanostructured silicon thermopiles by using silicon complementary metal-oxide-semiconductor (CMOS) technology, in a vertical device structure [69]. Using the same technology, a systematic increase in Γ_p from 25 to $84 \mu\text{W cm}^{-2} \text{K}^{-2}$ at room temperature was achieved in Si_{1-x}Ge_x ($x = 0.0, 0.01, 0.02$, and 0.03) blades as shown in table 1. These Γ_p s are comparable to the Bi-Te-based exotic μ -devices from the viewpoint of power output [103, 104]. This success is credited to the ability of the cost-effective CMOS technology that offers (a) a high-density integration of thermocouples, (b) low electrical and thermal contact resistances, and (c) the ability to tune electrical and thermal properties by controlling doped levels, the width of the thermopiles, and by engineering packing fraction [42, 69]. All these features make CMOS technology a more realistic technological solution for device

fabrication for powering IoT-based SNWs from ambient waste heat. Further tests of μ -TEGs based on silicon-NWs might provide interesting results as they are predicted to generate P_A in the range of $\sim \text{mW cm}^{-2}$ with CMOS technology using miniaturization and integration [105].

2.1.3. Outlook

Focus on the cost-effective fabrication processes such as CMOS, MEMS technology, which allow fabrication of devices with small footprint area and high power output, is increasing. More attention must be paid to the design of heat from the hot to the cold end propagation in the micro-TEGs, as the large temperature gradients will result in larger power output. As the key compatible materials for such technologies are mainly Si and SiGe, therefore, the discovery of more silicon compatible materials is also necessary. Beyond the need to deploy novel technologies replacing the standard ones some factors must be considered that currently hamper a larger utilization of thermal energy harvesters: (a) improvement in materials (b) circuits consumption with extremely low power (c) TEG often requires additional power-conditioning electronics to be integrated with the harvesters, and (d) the cost factor should be minimized. Finally, TEGs are perhaps the best alternative when the use of other energy harvesting technologies is impractical.

2.2. Piezoelectric energy harvester

Piezoelectric energy harvesters (PEH) convert mechanical vibrations into electrical energy using piezoelectric materials. PEH is a crucial technology for self-power sources of unattended electronics, WSN, and biomedical and wearable devices. Detailed literature about PEH has been released in the last decade [119–124]. Here, we focus on the fundamental working mechanisms of PEH, the key parameters of mainly used piezoelectric materials, and the design principle, given by their technological potential in the IoT field.

PEH is based on the direct piezoelectric effect, where piezoelectric materials generate electrical energy when under external stress. The effect is described by the constitutive equations [119]:

$$D_i = \varepsilon_{ij}^S S_j + d_{ij} \sigma_j, \quad (2)$$

where vector D and E are the electrical displacements, d_{ij} is the piezoelectric constant, S_j and σ_j are the mechanical strain and stress, ε_{ij}^S is dielectric constant under constant strain. For the converse piezoelectric effect, an electrical stimulus is converted into strain in the materials. With a similar equation:

$$S_j = c_{ij}^E \sigma_j + d_{ij} E_i, \quad (3)$$

where c_{ij}^E is mechanical stiffness under a constant electrical field.

For an efficient PEH, piezoelectric materials require a large electromechanical coupling factor (k), piezoelectric constant d . Notably, the k characterizes the mechanical–electrical energy conversion efficiency. For the transversal and longitudinal effects, e.g. the 31- and 33-mode, it follows [122]:

$$k_{ij}^2 = \frac{W_{\text{electrical}}}{W_{\text{mechanical}}} = \frac{e_{ij}^2}{\varepsilon_{ij}^T c_{pq}^E} = \frac{e_{ij}^2}{\varepsilon_{ij}^S c_{pq}^E + e_{iq}^2}, \quad (4)$$

where Y is Young's modulus, e_{ij} is the piezoelectric constant, ε_{ij}^T is dielectric constant under constant stress. For ambient mechanical vibrations, the piezoelectric energy harvester generally operates at much lower frequencies than the mechanical resonance of the materials. In such a case, the PEH operates in off-resonance conditions and the piezoelectric element can thus be approximated as a parallel plate capacitor, where the electric energy is given as $U = 1/2 CV^2$, or by the energy per unit volume [125]:

$$u = \frac{1}{2} (d \times g) (F/A)^2, \quad (5)$$

where C is the capacitance, V is the output voltage, F is the applied force, g is the piezoelectric voltage constant ($=d/\varepsilon_0 \varepsilon_r$), and A is the area. The product of d and g ($d \times g$) is the so-called FOM. With a given area and thickness under the same driving force, piezoelectric materials with a large FOM provide more power.

However, to maximize the mechanical harvesting, resonant devices are the choice. For maximum harvested electrical power in resonant conditions with high efficiency (η), want piezoelectric with high k and high mechanical quality factor Q_m :

$$A = \frac{\left(\frac{1}{2}\right) \frac{k^2}{1-k^2}}{\frac{1}{Q_m} + \frac{1}{2} \frac{k^2}{1-k^2}}, \quad (6)$$

where $Q_m = F_r/BW$, characterizes the sharpness of the piezoelectric's electromechanical response spectrum. BW is the bandwidth, denoting the frequency span around the resonance frequency (F_r). A high Q_m is important for resonant devices to avoid losses to heat and problems with self-heating.

Table 2. The piezoelectric and electromechanical properties of typical piezoelectric materials.

Material	Material type	k_{31}/k_{33}	d_{31}/d_{33} (pC/N)	g_{31}/g_{33} (10^{-3} Vm/N)	FOM ($\times 10^{-12}$)	Y (Gpa)	Reference
PVDF	Polymer	0.11/0.15–0.25	–23/33	–216/330	10.89	~3.0	[123, 139]
PMN-PT	Single crystal	0.76/0.94	–1283/2365	–21.22/39.11	92.49	16.5	[139, 140]
PZN-PT	Single crystal	0.50/0.90	–970/2000	–21.0/44	88.00	8.2	[141, 142]
PZT-5H	Ceramic	0.39/0.75	–274/593	–9.11/19.7	11.68	55	[139, 143]
KNN	Single crystal	0.646*/0.827	–77/162	–32.6/68.5	11.10	—	[144]
BTO	Ceramic	24.4/53.9	–92.3/205	–5.8/12.3	2.52	—	[145]

2.2.1. Materials

Piezoelectric materials for PEH can be both inorganic and organic. Among the inorganic materials, ferroelectric materials are often the best choice as they have high FOM and their properties are tunable by both composition and microstructure. Some high-performing materials are lead-based materials such as $\text{PrZr}_x\text{Ti}_{1-x}$ (PZT), $\text{Pb}(\text{Mg}_{1/3}\text{Nb}_{2/3})\text{O}_3$ (PMN). However, lead is toxic and is restricted to be used due to environmental concern. Among various lead-free ferroelectrics, BaTiO_3 (BTO), $\text{K}_x\text{Na}_{1-x}\text{NbO}_3$ (KNN), and solid solution of BTO with relaxor $(\text{Na}_{1/2}\text{Bi}_{1/2})\text{TiO}_3$ (NBT) show promising piezoelectric properties and attracting increasing interest in recent years.

In recent years, nanostructuring, doping, and defect chemistry are very effective strategies to boost the properties and performances of the piezoelectric material [126–128]. For ferroelectric, for instance, structuring of nano-domains can lead to superior properties with piezoelectric coefficients above 1000 pC N^{-1} [126]. Atomistic mechanisms causing such changes are not completely clarified and new tools in materials combinatorial screening, design, and simulation can be of great relevance in the coming years [129–131]. The role of interface and nanoscale also introduced new possible criteria for an extremely miniaturized system that requires high energy density [132–134].

For organic materials, phase polyvinylidene fluoride (PVDF) is one of the most common choices due to its overall performance. Polymeric materials are softer than ceramic piezoelectric (lower Y), and they have a negative strain with the electric field. Many efforts have been spent in the last decade to fabricate hybrid materials with synergic performances between mechanical flexibility, and easy processing of the polymers, with high piezoelectric performances of inorganic materials [135, 136]. However, such an integration is not always trivial as it requires novel methodologies and chemical methods to interface different materials efficiently. Recent advances in the field of nano-generators, however, indicate that novel heterostructures and carbon-metal oxide piezoelectric can achieve impressive results [137, 138]. Table 2, summaries such parameters for typical piezoelectric materials. Compare with inorganic piezoelectrics, organic materials such as PVDF has lower k and d values. However, the FOM value is comparable to PZT-based ceramics due to its large g value. Remarkably, PMN-PT and PZN-PT single crystals show much higher FOM compared with other materials.

2.2.2. Devices

Besides the selection of the material, the specific design of the PEH depends on its application requirement, such as the dimension limits, the mechanical energy source, the operating frequency, etc. Typical applications for WSN lay in the mW power generations [153]. However, the general trend is to reduce the power consumption of the sensor and communication systems, and PEH is the nW power generation, i.e. nano-generators, conveys a large variety of designs and solutions.

For the micro-power, extensive works have been carried out to improve the design of micrometric PEH in MEMS. Typical structures for silicon-based MEMS use a cantilever configuration with unimorph, bimorph, and a proof mass attached to the free end of the cantilever. The mass is used for tuning the resonance in the low-frequency range ($< \text{kHz}$) [146]. To amplify the effective piezoelectric strain of the piezoelectric materials, the cymbal structure with metal end caps is designed. However, the resonance frequency of this configuration is very high, and a high mechanical source is needed to activate the PEH with this configuration [147]. Another approach is the multilayer design, which stacks piezoelectric layers together to enhance the output [148]. In particular, Xu *et al* [149] designed a new structure, which combined the cantilever and cymbal structures. This configure uses the bending motion of the cantilever to compress the two cymbals to generate electrical energy. Therefore, the resonance frequency decreases effectively, making it suitable for low-frequency applications.

In recent years, hybrid energy harvesters (HEH), merging piezoelectric harvesting with other mechanical harvesting technologies, are attracting increasing research interest. An example of HEH combining piezoelectric and electromagnetic energy harvesting approaches is designed by Li *et al* [150] The authors

Table 3. Comparison of various piezoelectric energy harvesting devices. The main properties are characterized by the dimensions of the devices, power (U), output voltage (V), loading resistance (R), and operating frequency (f).

Material	Configuration	Dimension	U	R	V	f	Reference
PZT-5A ceramics	Cantilever	1 mm ³	375 μ W	200–300 k Ω	69.8 V	120 Hz	[146]
PMN-PT single crystal	Cymbal	$\approx 30 \times 5$ mm ²	14 mW	74 k Ω	45.7 V	500 Hz	[147]
Polypropylene (IXPP)	Stacked or/and folded IXPP Piezoelectret films	20 \times 20 mm ²	82 μ W	93 M Ω	—	400 Hz	[148]
PMN-PT single crystal	Cantilever + Cymbal	$\approx 35 \times 5 \times 8$ mm ³	3.7 mW	251 k Ω	38 V	102 Hz	[149]
PZT ceramics	Piezoelectric cantilever + electromagnetic coil	$\approx 60 \times 15 \times 40$ mm ³	3.0 mW	170 k Ω /24 k Ω ^a	—	73 Hz	[150]
PVDF	Hybrid cell/multilayered planar structure	63.5 cm ²	0.95 mW	140 M Ω	—	4.4 Hz	[151]
PZT ceramics	Piezoelectric bimorphs	47 \times 20 \times 0.5 mm ³	613 μ W	20 k Ω	13 V	20 r min ^{−1}	[152]

^a For piezoelectric and electromagnetic load, respectively.

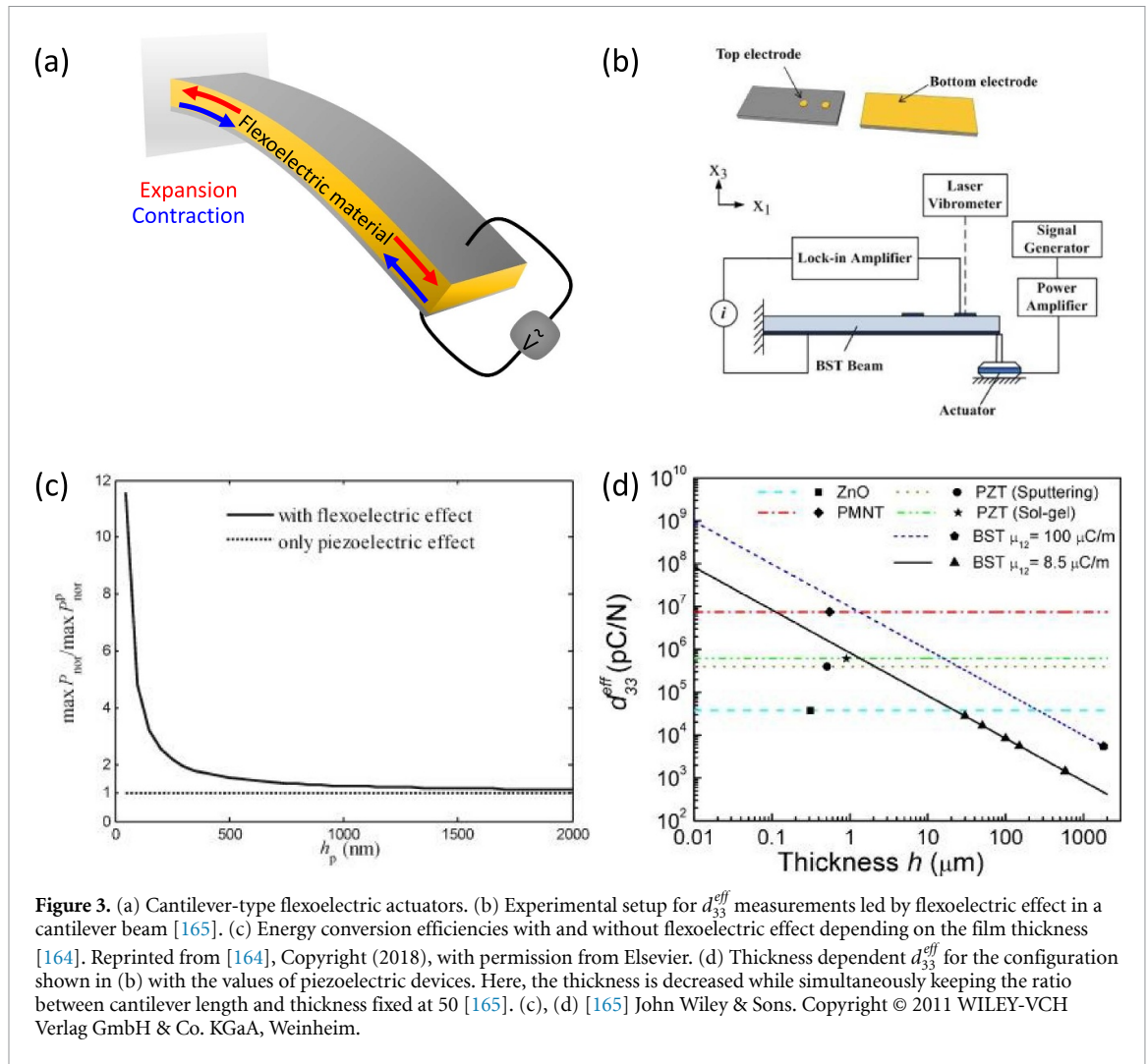
found that enhancing the coupling between the piezoelectric and electromagnetic coupling not only improves the power and power spectral density but also broadens the frequency range to capture random vibrations. However, the electromagnetic system can interfere with electronics and is generally avoided in IoT applications. Alternatively, Zi *et al* [151] designed a triboelectric-pyroelectric-piezoelectric hybrid cell composed of a sliding mode (TEG) and a pyroelectric-piezoelectric nanogenerator for hybrid energy harvesting. At a sliding frequency of 4.4 Hz, the TENG alone generates a power density of 0.15 W m^{−2}. Using the mechanical energy and friction-induced heat on the TENG, the hybrid device gives rise to a power density two times that generated by TENG alone. With a small load resistance of 1 k Ω , the energy efficiency could reach up to 26.2% with the potential to be further improved with proper operations or power management. Another promising application area of PEH is to capture the energy from fluid flow (e.g. wind, tide, etc). For example, Yang *et al* [152] designed a rotational piezoelectric wind energy harvester using impact-induced resonance. The optimal DC output power reaches 613 μ W across the 20 k Ω resistors at a rotation speed of 200 r min^{−1}. This kind of PEH is suitable for environments that are lacking vibrational energy but are rich in fluid flow (e.g. wind, water, etc). Table 3 summarizes the main properties of various piezoelectric energy harvesting devices.

In the field of nano generations, a wide variety of sensors are based on nano-tube, graphene, etc. For tribological energy, fluid flow, etc, many include biocompatible and integration on implantable and wearable flexible electronics. Although the power of such devices is currently too low, more efficient generation and reduced consumption of sensor and communication devices can be the next frontier for the IoT.

2.2.3. Summary and outlook

In summary, PEH are of particular interest as autonomous power sources. To further improve the overall performance, future works can be carried out via three different approaches:

- Optimize the electromechanical properties of the piezoelectric materials. Except for k , $d(e)$, other parameters are also relevant, such as mechanical quality factor Q , dielectric loss; the development of PEH based on lead-free piezoelectric materials, which is desirable for biomedical applications; the development of flexible piezoelectric materials, etc;
- Use of non-toxic materials. The most used piezoelectric materials currently in use are lead-based. However, new regulations in the act limit or even ban their use in future technologies. New materials with enhanced properties are thus needed and demonstrated for IoT applications, especially for those devices that are meant to be spread in the environment or in contact with the human body.
- The use of nanostructured and materials discovery initiatives would allow to further improve the performance of piezoelectric materials, and expand their application field, e.g. in the high-temperature range, where they are currently only a few compositions are in play.
- The design of PEH with novel configurations to tune the resonance frequencies for specific applications, broaden the bandwidth, the energy conversion efficiency;
- Improve the matching circuit and impedances of the electrical devices to reduce the electrical energy loss during transportation and storage.



2.3. Flexoelectricity for IoT applications

Piezoelectricity is traditionally used for converting kinetic energy into electricity in IoT devices., as the devices shrink and the active materials for harvesting reach the nanoscale, electromechanical conversion due to flexoelectricity can be comparable to or even more efficient than that from piezoelectricity. Flexoelectricity manifests itself as an electromechanical coupling between a strain gradient ($\partial \varepsilon_{kl} / \partial x_j$) and induced electrical polarization (P_i):

$$P_i = \mu_{ijkl} \frac{\partial \varepsilon_{kl}}{\partial x_j}, \quad (7)$$

where μ_{ijkl} denotes the flexoelectric coefficient [154, 155]. In analogy with piezoelectric energy conversion, where strain causes a change in polarization and induces a current between two electrodes as an attempt to screen the polarization, flexoelectric energy harvesters use strain gradients to induce polarization and current. Since flexoelectricity is a universal phenomenon exhibited in all-dielectric materials, the range of available materials for flexoelectric devices is much wider than that of piezoelectric devices, which is allowed only in 21 symmetry point groups. Consequently, not only harmful lead-composite materials can be avoided, but also extreme operating temperatures are possible due to the lack of Curie temperature [156]. Furthermore, flexoelectric devices do not need an elastic passive layer to be bent [157] (figures 3(a) and (b)), and do not suffer from the hysteretic nature of the spontaneous polarization that can induce a dramatic performance reduction and irreproducibility.

As shown in equation (7), flexoelectric polarization is directly proportional to the flexoelectric coefficient and strain gradient. Since the strain gradient scales inversely with the dimension of the material, the flexoelectric effect can dominate at the nanoscale where large values of the strain gradients up to $\sim 10^6$ – 10^8 m^{-1} can be achieved [158]. In this range of high strain gradients, unusual flexoelectric phenomena such as polarization enhancements [159, 160], flexoelectric field effects on electronic conduction [161, 162], and enhanced electromechanics [163] have been reported. In the same way, the

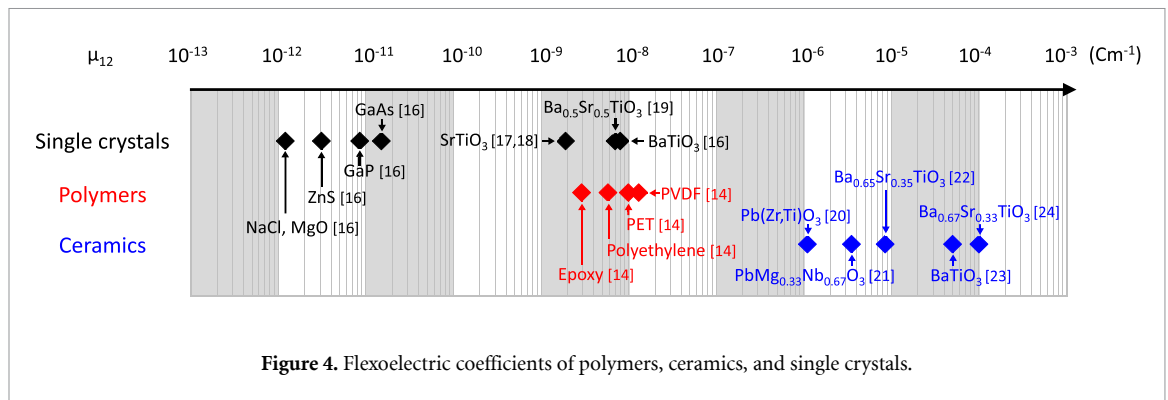


Figure 4. Flexoelectric coefficients of polymers, ceramics, and single crystals.

electromechanical properties of flexoelectric devices, such as energy conversion efficiency from mechanical energy to electrical energy, are highly enhanced at the nanoscale [164] (figure 3(c)). In several studies on flexoelectric cantilever devices, the effective piezoelectric coefficient (d_{33}^{eff}) has been considered to characterize the electromechanical properties of the flexoelectric cantilevers, which is defined as the generated flexoelectric charge per unit area by an applied normal force. The expression is

$$d_{33}^{eff} = \frac{6\mu_{3311}l^2}{Et^3}, \quad (8)$$

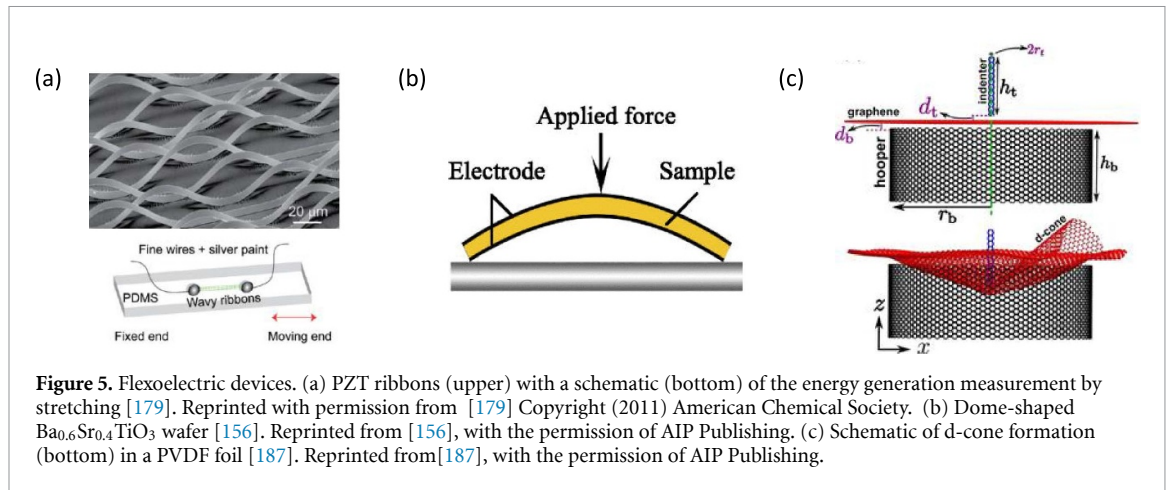
where μ_{3311} ($=\mu_{12}$ in a cubic symmetry, which is expressed in Voigt notation) is the transverse flexoelectric coefficient, l is the cantilever length, t is the cantilever thickness, and E is Young's modulus [165, 166]. Since d_{33}^{eff} is inversely proportional to the cube of the cantilever thickness, the electromechanical performance is significantly enhanced when the thickness decreases. The value of d_{33}^{eff} can be experimentally obtained by measuring the induced current flowing between a bottom and top electrode to screen the polarization in a vibrating flexoelectric cantilever [165] (figure 3(b)). Based on the measured d_{33}^{eff} values in a $\text{Ba}_{0.65}\text{Sr}_{0.35}\text{TiO}_3$ cantilever, the electromechanical performance in a few nanometer thick cantilever is expected to exceed that of piezoelectric ones by up to two orders of magnitude, as evident from figure 3(d) where the thickness is decreased while keeping $l/t = 50$ fixed [165].

2.3.1. Materials

Since d_{33}^{eff} is directly proportional to μ_{3311}/E , materials with a high flexoelectric coefficient or low Young's modulus are favorable for flexoelectric energy harvesting. Accordingly, soft polymers and oxides with high flexoelectric coefficients are promising candidates for flexoelectric devices. Polymers such as polyethylene, epoxy, and polyvinylidene typically have elastic moduli two orders of magnitude lower than ceramics, which could induce considerable enhancement in d_{33}^{eff} [167, 168]. However, to date, the flexoelectric coefficients of polymers ($\sim 10^{-9}$ – 10^{-8} Cm^{-1}) are in general 2–5 orders lower than those of ceramics ($\sim 10^{-6}$ – 10^{-4} Cm^{-1}) [165, 167, 169–175] (figure 4), although the large flexoelectric coefficient of α -PVDF ($\sim 10^{-5}$ Cm^{-1}) was reported, which is still controversial due to the discrepancy between the different authors with unclearness [167, 176–178]. Consequently, the flexoelectric coefficient and elastic modulus compensate each other, so the d_{33}^{eff} of polymers generally becomes smaller than for ceramics. Nonetheless, the stability of polymers under large deformations is better due to their softness, leading to a longer lifetime. The advantage of ceramics is the high flexoelectric coefficient, with for example μ_{3311} of paraelectric $\text{Ba}_{0.67}\text{Sr}_{0.33}\text{TiO}_3$ reaching $\sim 10^{-4}$ Cm^{-1} at room temperature, which is at least four orders higher than polymers and single crystals [175] (figure 4). In general, since the flexoelectric coefficient is proportional to the dielectric constant, materials such as $(\text{Ba},\text{Sr})\text{TiO}_3$, $(\text{Pb},\text{Sr})\text{TiO}_3$, and $\text{Pb}(\text{Mg}_{1/3}\text{Nb}_{2/3})\text{O}_3$ - PbTiO_3 (PMN-PT) have very large flexoelectric coefficients near the ferroelectric-to-paraelectric phase transitions, which could be manipulated for flexoelectric devices (figure 4) [154, 175, 178].

2.3.2. Devices

Experiments using various designs of flexoelectric devices including cantilevers have been successfully executed as shown in figure 5 and table 4 [156, 157, 165, 179, 180]. In the ribbon-type PZT devices, the piezoelectric effect is increased by 70% in the curved region, which seems to come from the flexoelectric effect due to the strain gradient effect formed when the curved structure is stretched (figure 5(a)) [179]. Periodic stretching (8% strain) of the device consisting of ten PZT ribbons generates an electrical current of ~ 40 pA by converting mechanical energy into electrical energy. In another device with a dome-shaped ceramic wafer made of $\text{Ba}_{0.6}\text{Sr}_{0.4}\text{TiO}_3$, a finite d_{33}^{eff} exists above the Curie temperature due to the flexoelectric



effect (figure 5(b)) [156, 181]. Using polymers, electrical energy can also be generated through flexoelectricity by a so-called d-cone formation in the wrinkle of a PVDF foil for wearable electronics (figure 5(c)) [180, 182].

Theoretical studies have also been performed to describe and optimize the electrical power analytically or numerically for the various designs of flexoelectric devices. In the cantilever design, induced electrical power is enhanced by 100% when both flexoelectricity and piezoelectricity are considered on nanometer scales compared to piezoelectricity alone [183]. It also has been reported that the output power of a cantilever depends on the residual surface effects such as surface elasticity, surface stress, and surface piezoelectricity [184]. The flexoelectric devices can be improved by considering a three-layered cantilever resulting in an energy conversion efficiency much larger than for single- or double-layered cantilever [185], or by attaching heavier masses at the end of the cantilevers arranged in an array to broaden the bandwidth of the resonance frequency [186]. It has also been investigated whether it is more convenient to connect the cantilevers in series or in parallel, which led to the conclusion that the induced power is largest for an array of cantilevers connected in series [186].

2.3.3. Outlook

Although the various experiments and theoretical analyses of flexoelectric devices have been studied, few papers report on experimental nanoscale devices [157, 188]. Among the reported results, the nanometer-thick SrTiO_3 cantilever is one of the most promising electromechanical devices for nanoscale flexoelectric energy harvesting [157]. In this device, not only the measured electromechanical performance is comparable to those of piezoelectric materials, but also its flexoelectric coefficient is maintained on the nanoscale with a similar value as in the bulk. This result is promising for making highly efficient flexoelectric devices from the recently developed freestanding membranes and cantilevers using oxide materials with thicknesses down to a single unit cell [157, 189]. In particular, a recent study reports an abnormally large flexoelectric polarization in the few-nanometer-thick epitaxial film with an extreme strain gradient value of $\sim 10^7 \text{ m}^{-1}$. This phenomenon is expected to originate from the interplay between piezoelectricity and flexoelectricity in the highly curved film with the unusual Poisson's ratio distribution along the film-normal axis [190]. By combining nanofabrication technology with this polarization enhancement in the extreme deformation range, 2D-like flexoelectric energy harvesters could be a promising route for efficient power generators in IoT devices.

2.4. Electromagnetic vibration energy harvester for IoT applications

Almost all large-scale power generation today consists of turning mechanical energy into electrical energy through electromagnetic radiation. It is also possible to utilize this phenomenon to harvest power for IoT devices. In this case, the mechanical energy is usually provided as the vibration of the environment on which the IoT devices are mounted. There are two approaches to harvesting vibrational energy, namely piezoelectric, and electromagnetic energy harvester (EMEH). In this section, we consider the latter approach. A few comparisons of these two types of technologies exist, but a single study on wearable devices finds that the efficiency of piezoelectric generators is quite lower, as is their power density, compared to magnetic harvesters [193].

The core principle of an EMEH is that ambient vibrations cause a permanent magnet to vibrate through the coil, i.e. a time-varying magnetic flux density (φ_B) through the coil which results in an induced electromotive force as $\mathcal{E} = -(\text{d}\varphi_B/\text{d}t)$.

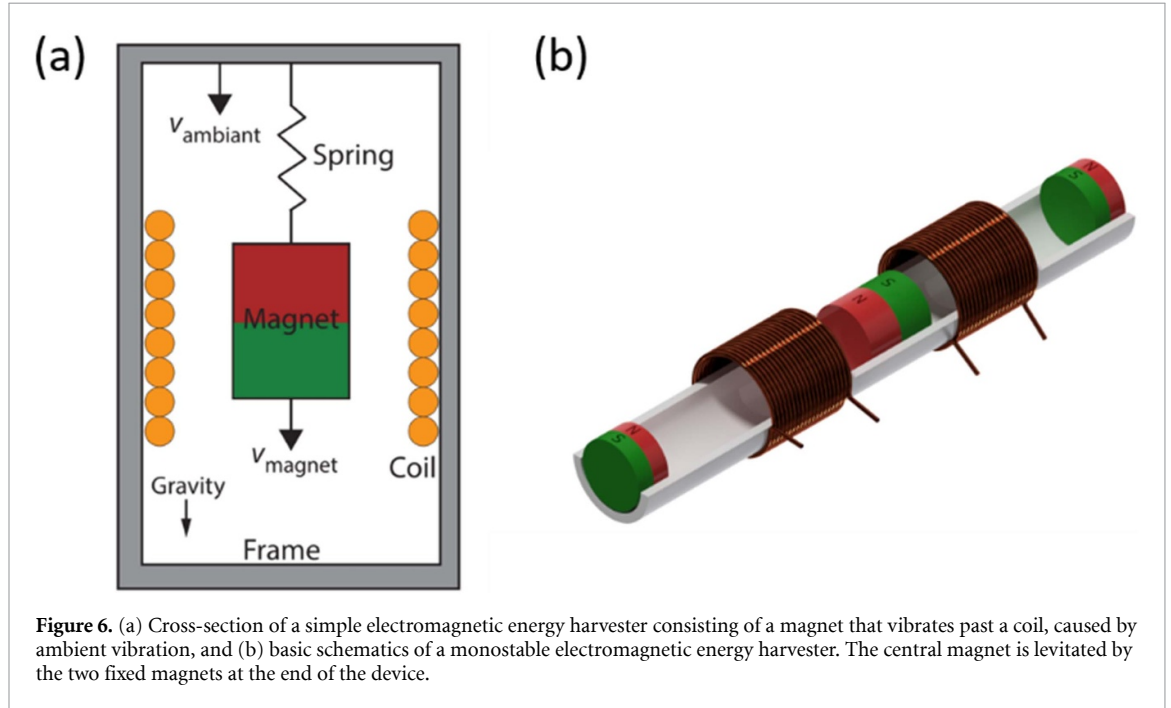
Table 4. Experimental and theoretical performances of flexoelectric devices (see table footnote^a).

Technology	Device geometry	Active material	Energy source	Input	Energy production	Output	Conversion efficiency	Operation temperature	Remarks
Flexoelectricity	Cantilever	SrTiO ₃	Elec.	Voltage	Mech.	Curvature/E-field = $3.33 \times 10^6 \text{ V}^{-1}$	—	RT	Exp.[157]
Flexoelectricity	Cantilever	Ba _{0.65} Sr _{0.35} TiO ₃	Mech.	Point force	Elec.	$d_{33}^{eff} = \sim 10^{-2} - 10^{-4} \text{ pC/N}$	—	RT	Exp.[165]
Flexoelectricity	Buckled ribbons	PZT	Mech.	Stretch (8% strain)	Elec.	$I_{SC \text{ Max.}} = \sim 40 \text{ pA}$	—	RT	Exp.[179]
Piezoelectricity									
Flexoelectricity	Domed wafer	Ba _{0.6} Sr _{0.4} TiO ₃	Mech.	Point force	Elec.	$d_{33}^{eff} = \sim 1 - 24 \text{ pC/N}$	—	RT—200 °C	Exp.[156]
Piezoelectricity		Na _{0.5} Bi _{0.5} TiO ₃ -BaTiO ₃				$d_{33}^{eff} = \sim 10 - 24 \text{ pC/N}$	—	RT—400 °C	
Flexoelectricity	Crumpled foil	PVDF	Mech.	Point force	Elec.	$I_{SC \text{ Max.}} = \sim 100 \text{ nA}$	—	RT	Exp.[180]
Piezoelectricity						$V_{OC \text{ Max.}} = \sim 0.06 \text{ V}$	—		
Flexoelectricity	Cantilever	PZT	Mech.	Vibration excitation	Elec.	Up to $\sim 1.8 \times 10^8 \text{ W m}^{-3}$	—	RT	Theor.[184]
Piezoelectricity									
Flexoelectricity	Cantilever	PVDF	Mech.	Vibration excitation	Elec.	—	Up to ~5%	RT	Theor.[185]
Flexoelectricity									
Piezoelectricity	Cantilever	Ba _{0.6} Sr _{0.4} TiO ₃	Mech.	Vibration excitation	Elec.	$> 0.04 \mu\text{W}$ (90–110 Hz)	—	RT	Theor.[191]
Flexoelectricity									
Flexoelectricity	Cantilever	Ba _{0.6} Sr _{0.4} TiO ₃	Mech.	Vibration excitation	Elec.	Up to $\sim 12.2 \mu\text{W}$	—	RT	Theor.[192]
Flexoelectricity									
Flexoelectricity	Circular plate	PVDF	Mech.	Vibration excitation	Elec.	—	Up to 6.6%	RT	Theor.[164]
Piezoelectricity									
Flexoelectricity	Circular membrane	PVDF	Mech.	Vibration excitation	Elec.	Up to $\sim 8 \times 10^8 \text{ W m}^{-3}$	Up to 11.93%	RT	Theor.[192]
Flexoelectricity	Ring	PVDF	Mech.	Vibration excitation	Elec.	Up to $\sim 50 \mu\text{W}$	—	RT	Theor.[192]

^a Elec.: Electrical energy, Mech.: Mechanical energy, d_{33}^{eff} : Effective piezoelectric coefficient, RT: Room temperature, $I_{SC \text{ Max.}}$: Short circuit maximum current, $V_{OC \text{ Max.}}$: Open circuit maximum voltage, Exp.: Experimental result, Theor.: Theoretical result.

Table 5. Power output from typical electromagnetic vibrational energy harvesting devices.

EMEH Motion	Category	Peak power density ($\mu\text{W cm}^{-3}$)	References
1D	Single coil single magnet	7229	[194]
	Single coil multiple magnets	8015	[195]
	Multiple coils single magnet	1710	[196]
	Multiple coils multiple magnets	2800	[197]
2D	Multiple coils multiple magnets	788	[198]

**Figure 6.** (a) Cross-section of a simple electromagnetic energy harvester consisting of a magnet that vibrates past a coil, caused by ambient vibration, and (b) basic schematics of a monostable electromagnetic energy harvester. The central magnet is levitated by the two fixed magnets at the end of the device.

2.4.1. Materials

To make EMEHs having as high power-density as possible, Neodymium Iron Boron or Samarium Cobalt magnets (also known as NdFeB and SmCo magnets respectively) are the most powerful type of permanent magnets commercially available today, with magnetic properties that far exceed AlNiCo and ferrite magnet materials. While NdFeB magnets have a high-magnetic remanence and much higher coercivity than other permanent magnets, SmCo magnets rank similarly in strength, but they have much higher temperature stability. Both types have been used in manufacturing EMEHs with different power densities for four categories of one-dimensional harvesters and one category for two-dimensional motion as described in table 5.

2.4.2. Device

The simplest EMEH one can consider is a permanent magnet mounted on a spring and surrounded by a coil [199], as illustrated in figure 6(a). When ambient vibrations cause the spring to oscillate, the permanent magnet moves through the coil, causing a time-varying magnetic flux density and thus an induced electromotive force.

However, such magnets-on-a-spring devices can be plagued by mechanical friction, and also the wear and tear of the spring. Further, making these devices strong enough to act as harvesters for IoT devices would likely be problematic. Instead, the focus on electromagnetic energy harvesting has been on utilizing a configuration of permanent magnets that levitates a single permanent magnet with a spring-like restoring force. Such a device typically consists of three coaxial cylindrical permanent magnets, facing each other with polarities that make the middle magnet keep levitating, avoiding mechanical friction [196, 200, 201], as shown in figure 6(b). However, stable magnetic levitation is not possible as per Earnshaw's theorem, the permanent magnet must be enclosed in a cube, to prevent the middle magnet from flipping. We here term such configurations as mono-stable. When subjected to an ambient vibration the middle levitated magnet moves through one or more coils where an electromotive force is induced.

A substantial number of such EMEH mono-stable harvesters have been realized and tested. These can employ block or ring magnets, circular or rectangular containers, a different number of coil windings,

guidance for the middle magnet with or without spacers, and the spring on the top and/or the bottom edges to replace the fixed magnets [202]. These devices can also be classified according to e.g. their excitation and robustness [203]. A good physical description of the mono-stable EMEH devices exists, as the force on the levitated magnet can often be approximated with a spring-like force [201]. There is generally a very good agreement between modeling and experimental data for these devices [203], e.g. a mean absolute percentage error of 6% has been demonstrated [204]. Most current-state devices are in the centimeter range and produce a voltage in the mV–V range and power in the mW range [202], although this depends on the exact way that the device is operated as EMEH devices can have a quite complex phase space [205].

Besides the mono-stable EMEH design, several other designs have been considered. For example, in one approach, the coil is considered as the moving part instead of the permanent magnet [206]. In this case, the magnetic flux is enhanced due to a steel frame, but a couple of stoppers are needed at the edges of the harvesters. Likewise, attaching a coil winding to a rod that is exposed to ambient vibrations and vibrating this through a double concentric Halbach array magnetic structure can also be used to harvest power [207]. It is also possible to modify the one-directional movement of the free magnet in an EMEH to a multi-dimensional EH by making the fixed magnet similar to a doughnut-shaped container [208] and such devices have recently been shown to have a larger and more easily tunable harvesting bandwidth compared to monostable EMEHs [209, 210]. Finally, permanent magnets and springs can also be combined [211, 212] to specifically harvest power from very low-frequency vibrations.

Other electromagnetic harvesters: other electromagnetic energy harvesting designs than the levitated magnet design and the variants of this discussed above are also possible. One example is a harvester with a spring-less spherical permanent magnet with a non-uniform mass distribution, that generates a roly-poly-like motion in response to external vibrations, which induces a current in a coil [213]. Other designs use a joint about which the floating magnet in figure 6(b) rotates [214] or more chaotic movements using the random movement of a soft composite in a magnetic field with vibrations [215]. Moreover, Bi-stable prototypes have been also contemplated [216] so that moving restriction is attained in only one axis of the 3D space. The moving magnets keep levitating due to the repulsive force interaction with a set of permanent magnets around it. As a result, such EMEH can generate power from low frequencies and low g applications. Complex three-dimensional stable harvesters have also been studied [217], consisting of a set of lifting magnets, diamagnetic plates made of pyrolytic graphite, a floating magnet, and a couple of copper coils. This is also suitable for low frequency and low g applications. Finally, a design where the moving magnet does not depend on the repulsive force produced by a fixed magnet, but by a fluid has also been presented [218]. This EMEH employs a spherical magnet that rotates while fluid flows through the harvester.

2.4.3. Outlook

To be useful as energy harvesters for IoT applications, harvesters must typically be small e.g. about 1 cm^3 in size. This is a challenge for EMEHs, as these contain permanent magnets which have to be manually handled and positioned in the harvester and cannot be produced with e.g. lithography or other manufacturing techniques. However, microfabricating the coil in an EMEH is not challenging, and examples exist of harvesters where the common coil winding is replaced by a micro coil that can be fabricated on silicon or printed circuit boards [219], an example of such type of harvester can be seen in [220], where the micro-fabricated coil is placed on the bottom edge of the harvester. When the magnets are also miniaturized so that a miniaturized-sized device is produced, a few micro-watts of power can be generated when vibrated [221].

EMEHs have in the recent decade been shown to be able to produce a reasonable amount of power from the ambient vibrations, with power densities up to 8 mW cm^{-3} [195, 202], which is enough to supply energy to a sensor as it is detailed in the self-power monitoring system described in [210]. The technology is advantageous compared to other harvesting technologies because there is no wear on the device, the induction mechanism is simple and direct, and the device has low internal friction.

However, several issues must still be tackled with EMEH devices before these are fully ready to be commercialized. Besides the issue with miniaturization discussed above, further optimization of the design of the device, especially with a focus on multivariable performance optimization [198, 202] is necessary. Finally, ensuring a broader frequency harvesting range is likely needed for EMEH to be truly adaptable to the range of environments that IoT devices experience.

2.5. PV generators for IoT

PV convert light into electrical energy and is another promising way to power IoT devices [222–226]. IoT technology focuses mainly on indoor usages, where the spectra and intensities of light sources are quite different from the standard outdoor AM1.5G solar irradiation. In contrast to the AM1.5G solar spectrum,

which includes a significant portion of intensity in the infrared region, indoor light sources from fluorescent and light-emitting diodes (LEDs) have a spectrum with intensities primarily in the visible light range, from 350 to 700 nm, and with much weaker light intensities; the intensity from a ~ 500 lux white light-emitting diode (WLED) is only roughly 0.15% of the AM1.5G 1 sun intensity [31, 227]. Consequently, conventional PV based on e.g. crystalline silicon, which possess a narrow indirect band gap of 1.12 eV and being optimized for outdoor use, do not perform well under indoor lightning due to a non-ideal match between the absorption spectrum of the materials, and the illumination spectrum of the indoor light sources. This results in significant thermalization losses, and from that too low photovoltages. This has recently driven research towards the use of new thin-film PV technologies based on wider band gap semiconductors for IoT applications.

The performance of a PV device is characterized by the power conversion efficiency (PCE), which is defined as the fraction of the electrical output power to that of the incident light power [228]: $PCE = \frac{J_{sc}V_{oc}FF}{P_{in}}$, where J_{sc} , V_{oc} , FF , and P_{in} are the short-circuit current density, open-circuit voltage, fill factor, and input light power, respectively. The numerator in equation (1) thus represents the electrical output power of the solar cell. As noted above, both the light spectrum and intensities for indoor and outdoor (sunlight) light sources are different, which is affecting the P_{in} parameter. Depending on the illuminance of the indoor light source, an output power of $10\text{--}100 \mu\text{Wcm}^{-2}$ can typically be reached by indoor PV devices [222], which also means that a 10 cm^2 indoor PV device is large enough to power most wireless protocols used within IoT [7]. Further improvement of this requires optimization of materials and device architectures towards utilization under indoor light. Furthermore, such materials and architecture choices may affect not only efficiency but also the lifespan of the PV devices. An example of such considerations is for organic PV (OPV); for outdoor use, the PCE of OPV is slightly lacking behind that of silicon PV, which for indoor use is the opposite.

In terms of stability and lifespan, OPV still lacks behind silicon PV due to photo-degradation of organic molecules when exposed to light, heat, water, and oxygen [229]. In fact, the stability topic of organic photovoltaics (OPV) has been a large research field of its own for many years, also demonstrating the complexity of this topic due to the presence of multiple possible degradation routes. It is well known that photo-oxidation of organic molecules may take place under simultaneous exposure to oxygen and light [230, 231], something that can be partially mitigated via antioxidant additives [232, 233] and different encapsulation routes [234, 235]. The antioxidants are relevant as industrial processing always allows for inclusion of small amounts of oxygen in the cell during fabrication (done in air), which even the best encapsulant cannot protect against. Transport layer interface degradation is also well-known, in particular for many new non-fullerene based organic solar cell systems [236, 237], for example due to photocatalytic decomposition of these molecules at metal oxide interfaces. This has made studies on transport layer interfaces very relevant for the OPV field [238]. For indoor PV, however, these effects are much less of an issue due to the lack of harsh and alternating weather conditions, and in particular, the different light spectrum and intensity at indoor conditions, i.e. low light conditions and lack of UV component. Indeed, the photo-oxidation of OPV molecules is well-known to be wavelength dependent, and much more effective under UV light [239]. This leads to much more stable OPV devices at indoor conditions, which opens up for new routes for some emerging PV technologies within IoT applications.

2.5.1. Materials

The theoretical efficiency limit of a PV cell under a given incident illumination spectrum and intensity has been calculated for indoor light sources. For a single-junction solar cell with an optimum band gap lying between 1.82 and 1.96 eV, a theoretical PCE of higher than 50% can be reached [240], compared to the PCE limit of $\sim 33\%$ under AM1.5 G solar irradiation using a semiconductor with a band gap of 1.34 eV [241]. The effect of the semiconductor band gap on the PCE can be seen when comparing crystalline and amorphous Si [242]. While crystalline Si solar cells only deliver a PCE of $< 10\%$ under low-lux LED illumination, due to the low band gap of 1.12 eV, PV based on amorphous Si with a much wider band gap of ~ 2 eV have demonstrated PCEs of $> 15\%$ [242, 243]. Similarly, materials with band gaps that are normally not ideal for outdoor PVs can perform well under indoor lightning. For indoor OPV (1650 lux illumination) based on the polymer donor PM6 and the small molecule NFA Y6-O [244], an impressive PCE, of 30.9%, and an open-circuit voltage, V_{oc} , of 0.84 V has been recorded, mainly given by the strong light absorption between 400 and 800 nm for this material system. For indoor perovskite solar cells (iPsc), a PCE of 40% has been recently achieved for $(\text{FAPbI}_3)_{0.92}(\text{MAPbBr}_3)_{0.08}$ cells under 824 lux illumination [245]. Although many metal oxides are too wide band gap materials (above 3 eV), some binary and ternary oxide semiconductors exhibit strong optical absorption in the visible light range. For example, Cu_2O [246], $\alpha\text{-Fe}_2\text{O}_3$ [247, 248], $\text{Zn}_{0.5}\text{Mn}_{0.5}\text{O}$ [249], and $\text{Zn}_{1-x}\text{Cd}_x\text{O}$ ($x = 0.4\text{--}0.7$) [250–253] all have band gaps in the range of 1.8–2.2 eV and thus may also be promising materials as light absorbers for indoor PVs to power IoT devices.

Table 6. Overview of recent works in perovskite solar cells (adapted from [222]).

Layer structure	Active-area (cm ²)	Irradiance (Lux)	Light type	PCE (%)	Voc (V)	Jsc ($\mu\text{A cm}^{-2}$)	FF (%)	References
Glass/ITO/SnO ₂ /(MA _{0.91} FA _{0.09})Pb(I _{0.094} Br _{0.06})/HTM/Au	0.2	1000	LED (5000 K)	38.2	0.96	188	78.7	[259]
Au/poly(3-hexylthiophene) (P ₃ HT)/Poly[(9,9-diocetylfluorenyl-2,7-diyl)-co-(4,4'-(N-(4s butylphenol)diphenylamine)(TFB)/CsPbI ₂ Br/SnO ₂ /ITO/glass	0.1	1000	LED (6000 K)	27.16	0.94	122.0	0.77	[263]
Glass/ITO/SnO ₂ /Cs _{0.05} (FA _{0.6} MA _{0.4}) _{0.95} Pb(I _{0.6} Br _{0.4}) ₃ /PEAI/spiroOMeTAD/Au	0.2	1000	White LED (cool)	33.42	0.95	135	69.6	[264]
Glass/FTO/cTiO ₂ /(FAPbI ₃) _{0.97} (MAPbBr ₃) _{0.03} /spiroMeOTAD/Au	0.08	824.5	White LED (2700 K)	40.1	1	152	79.5	[245]
FTO/SnO ₂ /Cs _{0.05} (FA _{0.85} MA _{0.15}) _{0.95} Pb(I _{0.85} Br _{0.15}) ₃ /SpiroOMeTAD/M-oO ₃ /Ag	—	1000	LED	37.9	0.96	156	79	[262]
Glass/FTO/SnO ₂ /TiO ₂ /MAPbI ₃ /spiro-MeOTAD/Au	0.2	1000	LED	29.83	0.81	154.4	66.3	[265]
Glass/FTO/TiO ₂ /CsPbBrI ₂ :(NH ₄)C ₂ O ₄ /PTAA/MoO ₃ /Ag	—	1000	FL	28.48	0.75	170	62	[266]
Glass/FTO/SnO ₂ -PbO/MAPbI ₃ /spiro-OMeTAD/Au	0.24	285	—	34.2	—	—	—	[267]
Glass/FTO/TiO ₂ /CsPbBrI ₂ (NH ₄)C ₂ O ₄ /PTAA/MoO ₃ /Ag	—	1000	FL	28	0.75	170	62	[266]
Flexible glass/ITO/SnO ₂ /m-TiO ₂ /CH ₃ NH ₃ PbI ₃ /Spiro-OMETAD/Au	0.1	200	LED	20.6	0.82	30.7	72.2	[268]
ITO/SnO ₂ /Pe/PEABr/Spiro/Au	—	1000	LED	42.1	1.10	5810	77.8	[269]
Glass/FTO/SnO ₂ /perovskite/spiro-OMeTAD/MoO ₃ /Ag	—	1062	LED	31.9	0.92	0.150	77.1	[270]
CsPbI ₂ Br ₂ :Pb(Ac) ₂ /PM6	—	1000	LED	33.7	1.15	118	81.8	[271]

2.5.2. Devices

OPV have gained a lot of popularity due to their desirable properties, such as light-weight, mechanical flexibility, easy processability, and low-cost manufacturing, making them a cheaper alternative to the conventional Si-based solar cells. Owing to the switch from fullerene to non-fullerene acceptors (NFAs), the PCE of OPV has reached a record of 18.2% under 1 sun illumination (outdoor) [254], cutting it close to state-of-the-art Silicon-based PV, and above 30% under 1650 lux illumination (indoor) [244]. In contrast to fullerene acceptors, NFAs offer a larger degree of tunability in light absorption and also low voltage loss upon free charge carrier generation, which has been the main driver for this improvement. Besides their low environmental impact, OPVs are especially well-suited for indoor applications thanks to their large absorption coefficients, low leakage currents, and the tunable energy levels of organic semiconductors typically having band gap values well-suited for indoor lighting [255–257].

While the low light intensity generally causes lower open-circuit voltages for indoor applications, it also modifies the effect of charge recombination on device performance. Under low-intensity light, OPV has a higher susceptibility to leakage current and charge recombination caused by trap states, due to the overall lower charge carrier density [244, 256], which is in contrast to bimolecular recombination effects that are minor under these conditions [255]. These issues can be minimized by passivating defects and tuning the energy levels of the charge transport layers, to achieve a more desirable energy level alignment and form charge carrier selective contact layers, as well as by adjusting the energy levels of the organic active layer materials to increase the V_{oc} under low-intensity light [258]. Part of that is behind the success demonstrated recently for NFA OPV cells.

The integration of an unconventionally used copolymer has enabled the PCE of iPsc to reach the highest efficiency of Pb mixed halide iPsc of 38.2% [222, 259] under indoor 1000 lux. In tables 6 and 7, an overview of the most recent works on indoor perovskite and organic solar cells are summarized.

Table 7. Overview of recent works in organic solar cells.

Layer structure	Active area (cm ²)	Irradiance (Lux)	Light type	PCE (%)	Voc (V)	Jsc ($\mu\text{A cm}^{-2}$)	FF (%)	References
Glass/ITO/PEDOT/PM6:Y6-O/PDINO/Al	0.08	1650	LED (3000 K)	30.89	0.84	24.5	76	[244]
Glass/ITO/PEDOT:PSS/PM6:Y6-Y-Th2/PDI-NO/Al	0.00477	1000	LED	22.72	0.701	32.01	74.48	[272]
Glass/ITO/PEDOT:PSS/PBDB-TF:IO-4Cl/poly(PFN-Br)/Al	1	1000	LED (2700 K)	26.1	1.1	90.6	79.1	[273]
Glass/ITO/ZnO/PBDB-TSCL:IT-4 F/MoO _x /Ag	—	500	FL	21.53	0.63	60.44	76.29	[274]
PET/IMI/PEI/PC ₆₀ BM(443 nm)/PEDOT:PSS/Ag(Full Roll to Roll)	0.55	1000	LED	26	0.643	124	69.0	[275]
Glass/ITO/ZnO/D18:FCC-Cl/MoO ₃ /Ag	0.07	2000	LED (2600 K)	30.1	0.975	245.4	80.1	[276]
Glass/ITO/ZnO/PM6:FCC-Cl/MoO ₃ /Ag	0.07	2000	LED (2600 K)	28.5	0.914	244.1	81.2	[223]
ITO/ZnO/PFN-Br/PM6-BTP-BO-4Cl	0.04	1000	LED (3000)	20.20	0.67	104.82	73	[264]
ITP/ZnO/Polyethylenimine Ethoxylated/PBDB-T:M-ITIC-O-EH/MoO _x /Ag	0.1	1000	LED	21.6	678	163.9	69.8	[277]
PM6-BTP-BO-4CL	—	1000	LED	20.2	0.67	115.6	0.73	[278]
6K-PDMS/PEDOT/PSS	—	1500	LED	20.5	0.732	178.4	74.3	[279]
PBDB-TF:Y6-O	—	1650	LED	30.8	0.84	245	76	[244]

As mentioned above, the illumination intensities play an important role in the mechanisms governing the PCE of solar cells, such as current generation and recombination effects. Consequently, the PCE value is influenced according to the subjected illumination. Therefore, for further enhancement of PCE, the assessment of the traps associated with recombination mechanisms influencing the power efficiency must be considered both indoor and outdoor, especially under low light intensities, where the effect of these traps is prominent [222]. Attempting to mitigate the effects of trap-assisted recombination would require some treatments, for instance, the passivation of interface defects or the implementation of some additive to enhance their PCE [260]. Additionally, it was possible to fabricate a low-trap CsPbI_{2.7}Br_{0.3} perovskite cell that performed superbly and outperformed c-Si cells in low-light conditions under illumination from a 1000 lux fluorescent lamp and a WLED, the optimized device attained PCE values of 32.69 and 33.11% [261]. Besides, doping by strong electron acceptor of the perovskite demonstrated a viable method for effective charge transport, opening the door to their use as IPVs. The ideal perovskite solar cell has achieved an outstanding indoor PCE with PCE values of 37.9% under 1000 lux low light and 21.3% under one sun illumination [262]. This emphasizes the importance of gauging the trap states experimentally in indoor PV technology, in addition to a standard comparison of the solar cell performance between indoor light sources.

2.5.3. Outlook

Compared to conventional outdoor solar cells, indoor PV is still in an early stage. The effects of lower light intensity and hence reduced photo-generated carrier concentrations should be more carefully studied for device optimizations [222]. Currently, a standard protocol for indoor light sources does not exist [226]. We emphasize the importance of establishing such a standard for reporting efficiencies of indoor PVs, such that a fair comparison can be made.

While OPVs and perovskites have demonstrated high PCE values of >30% under indoor light illuminations, stability issues remain although they are less significant under indoor conditions. Therefore, the commercialization of indoor OPV has also started recently. Furthermore, perovskite solar cell efficiencies at low light illuminations have rendered them to be promising for indoor application compared to other technologies. For both OPV and perovskite, the potential low cost is a further driver for commercialization. At present time, efforts to implement these photovoltaic technologies in indoor applications, and more specifically, to power IoT devices, are already yielding promising results. For example, iOPVs have successfully been used to supply smart temperature and humidity sensors using indoor-like illuminations as low as 300 lux [256].

3. Section-II: energy storage and on-demand powering devices

In this section, we present recent developments on energy storage and on-demand power devices used for IoT applications. The major focus of this section is devoted to battery technologies which represent the most mature and most common solution for microscale powering devices (section 3.1), while we additionally discuss micro-fuel-cell-based technologies as an alternative pathway for on-demand powering of IoT devices, which may complement battery technologies in applications where fast charging/fueling processes are needed.

3.1. Batteries for IoT applications

IoT devices with energy harvesting may work as independent devices without onboard energy storage [280]. Most of these micro-harvesters are based on intermittent energy sources, therefore, integration of rechargeable energy storage in IoT devices is of importance for continuous, long-term, and independent functionality, ensuring that the harvested energy does not go to waste. Combining energy harvesting and storage allows utilization of very low-power energy sources while still enabling higher energy use at a given time independently of the availability of the external energy source [280]. Integrating energy harvesting and storage may lead to efficient utilization of the harvested energy, e.g. by combining PV with a lithium-ion battery (LIB) [281, 282].

The requirements for IoT rechargeable batteries are very dependent on their size and use. For instance, flexible batteries could be of interest for IoT wearables [283, 284], high- or low-temperature performance would be a need for devices working in extreme environments and micro-batteries would be needed for very small IoT sensor devices. Generally, the characteristics of IoT on-board rechargeable batteries are e.g. size, reliability, low self-discharge, safety, and durability. The batteries should ideally be scalable with a tailor-made design and format so that the batteries are made to fit the device instead of having to design the device to fit a specific battery format.

All-solid-state thin-film lithium batteries (TFLBs) with outstanding safety and excellent integration ability are the best candidate so far to fulfill this purpose [285]. Similar to LIBs, TFLBs consist of cathode, electrolyte, anode, and collectors [286]. This section presents the state-of-the-art materials for cathode, anode, and electrolyte with different advantages and challenges. Then, some full cell assembly is presented with some commercial examples. TFLBs are highly customizable, which could support low-power IoT devices with a power consumption of between $10\text{--}1000\ \mu\text{W cm}^{-2}$. Higher energy density and longevity are what is been pursued in the development of TFLBs. TFLBs can be fabricated by various vacuum-based and wet-chemical methods. For physical vapor deposition (PVD) vacuum-based methods, there are magnetron sputtering and pulsed laser deposition (PLD); atomic layer deposition (ALD) is a common technique in terms of chemical vapor deposition vacuum-based methods. Sol-gel, spray pyrolysis, dip coating, and spin coating are examples techniques of wet-chemical methods [287]. A schematic of the cross-section of a TFLB is shown in figure 7. In this review, we are focusing on PVD methods, which have better integrations of micro on-chip devices.

3.1.1. Materials

3.1.1.1. Electrolyte

Many materials have been developed as solid electrolytes (SEs) including inorganic, organic, and inorganic/organic hybrid materials [289, 290]. For thin-film batteries produced by deposition methods, PLD, sputtering or ALD, inorganic materials are the most relevant SEs. LiPON , $\text{Li}_{3-x}\text{PO}_{4-y}\text{N}_z$, is a lithium phosphate-derived material, which has been used extensively as SEs in thin-film LIBs despite its rather low intrinsic lithium-ion conductivity [290], due to its good stability in air and against metallic lithium [291]. NASICON-type (structurally equivalent to $\text{Na}_{1+x}\text{Zr}_2\text{P}_{3-x}\text{Si}_x\text{O}_{12}$) materials e.g. $\text{LiM}_2(\text{PO}_4)_3$ ($\text{M}=\text{Zr}, \text{Ti}, \text{Hf}, \text{Ge}$ or Sn) showed good lithium-ion conductivities [289], and partially substituted materials, e.g. $\text{Li}_{1.3}\text{Al}_{0.3}\text{Ti}_{1.7}(\text{PO}_4)_3$ (LATP), show even higher ion conductivity. The major issue of NASICON electrolytes is the interfacial instability both on cathode and anode [292]. Another perovskite-type material with a general formula of ABO_3 also has good Li-ion conductivity ($10^{-4}\ \text{S cm}^{-1}$). Lithium lanthanum titanate (LLTO) is one of the most promising perovskite electrolytes with a high ionic conductivity due to the vacancies in the A sites. Similar to LATP, the electrolyte/electrode interface of LLTO is thermodynamically unstable, resulting in increased resistance [293]. Garnet structure solid electrolyte (LLZO) is a promising material due to its high ionic conductivity and good chemical stability against metallic lithium. This outstanding stability is originated from the kinetic stabilization of the decomposition interphases [294]. However, it suffers from Li metal dendrites growth, due to the grain boundary originating from its polycrystalline structure [295]. The growth of this dendrite will eventually lead to a short-circuit, which hinders the use of Li metal as the anode material. Employment of single-crystal or amorphous LLZO could block the Li dendrite growth [295]. The

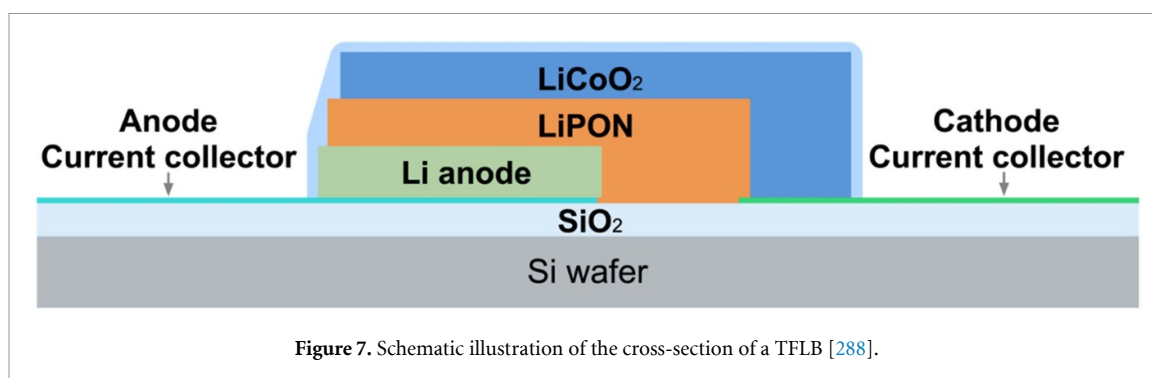


Figure 7. Schematic illustration of the cross-section of a TFLB [288].

ionic conductivity of thin-film LLZO is reported to be lower than that of bulk LLZO, due to the difficulty of depositing Li-containing thin films with sufficient crystallinity and density. This difficult crystallization of LLZO is caused by the volatilization of lithium at higher temperatures (both deposition and annealing) which results in unexpected stoichiometry [296]. More importantly, the higher surface-to-volume ratio of thin-film materials compared to bulk materials brings more severe Li loss [297]. Several sulfide-based SEs have been discovered with very high lithium-ion conductivity, especially related to $\text{Li}_{10}\text{GeP}_2\text{S}_{12}$ (LGPS). An example is $\text{Li}_{9.54}\text{Si}_{1.74}\text{P}_{1.44}\text{S}_{11.7}\text{Cl}_{0.3}$ (LSPS) [298], which has the highest room-temperature lithium ionic conductivity published so far. However, sulfide-based SEs such as LGPS and LSPS is not stable both mechanically and chemically against Li metal. They are also suffering from the toxic gas released when exposed to air [299].

3.1.1.2. Cathode

Current high-performance cathode materials used in thin-film micro-batteries are mostly lithium transition metal oxide materials. Due to the wider electrochemical window of SEs, some high-voltage cathode materials with promising specific capacities are readily available. LiCoO_2 (LCO) has an exceptionally high volumetric capacity (700 Wh l^{-1}) and relatively high gravimetric capacity [300]. The charging cut-off voltage of LCO is about 4.2 V, which gives a gravimetric capacity of about 140 mAh g^{-1} . Recently, the cut-off voltage had been pushed up to 4.55 V, which delivers a promising 190 mAh g^{-1} specific capacity [301]. High-voltage spinel $\text{LiNi}_{0.5}\text{Mn}_{1.5}\text{O}_4$ (LNMO) has been intensively studied [302, 303] and employed in TFLBs [302–305]. It can reach a high voltage of up to 4.7 V [306]. The specific capacity of PLD deposited LNMO is significantly lower than that of bulk LNMO, due to the difficulties of maintaining its ordered P4_332 structure at 700°C , which is the normal temperature for PLD and sputtering [307]. Ni-rich layered oxide (NMC or NCA) is commercially used in electric vehicles, which is a promising high-capacity material with reduced Co content. The synthesis of thin-film Ni-rich material with good crystallinity is challenging. Only a few studies had reported Ni-rich TFLBs [308]. Li-rich layer-structured oxide (LRO) materials could reach a capacity of 275 mAh g^{-1} with a 4.9 V cutoff voltage, which is higher compared to other cathode materials discussed here [309]. However, the lithium loss during the deposition and annealing processes is still a challenge, especially at a higher temperature [310, 311]. The stoichiometry of LRO is also difficult to control using vacuum-based deposition methods [311]. Low initial Coulombic efficiency, and poor cycle and rate capability also hinder the implementation of LRO [312].

3.1.1.3. Anode

$\text{Li}_4\text{Ti}_5\text{O}_{12}$ (LTO) is a Li-insertion material that is widely employed as an anode material in TFLBs. The main advantage of LTO is that the volume exchange during charge/discharge is minimal ($<1\%$), so-called zero-strain materials [313]. However, the low capacity ($<175 \text{ mAh g}^{-1}$) and high voltage (1.55 V vs Li) make it less appealing compared with silicon ($>3500 \text{ mAh g}^{-1}$) and lithium metal (3860 mAh g^{-1}) [314–316]. The high capacity and good lithography process compatibility of Si make it a promising candidate for on-chip TFLBs. The main challenge of Si used as an anode material is the large volume expansion ($>400\%$) [317]. The large volume change may cause a structural failure which leads to a short lifespan. To relieve the effects of this drastic Si volume change, many 3D-structured silicon anodes have been designed [318, 319]. This high-aspect ratio of 3D Si anode could also increase the areal power density of TFLBs. Li metal anode has the highest gravimetric capacity among all anode materials. The utilization of Li metal anode could drastically increase the gravimetric capacity of Li batteries. However, the lithium dendrite/void formation during the discharge (plating)/charge (stripping) process gives a critical current density which limits the high power application of Li metal anode [320]. Dendrite formation is less crucial for TFLBs when employing LiPON or amorphous LLZO as SEs, because these homogeneous defects free materials can completely prevent the dendrite penetration [295, 320], for instance, a study using LiPON as SE, $(\text{LiNi}_{0.5}\text{Mn}_{1.5}\text{O}_4)$ LNMO as cathode

Table 8. Examples of commercial TFLBs.

Manufacturer	Model	Composition	Capacity (μAh)	Size (mm)	Weight (mg)	Cycle life	Operating temperatures
TDK Electronics AG	CeraCharge [326]	$\text{Li}_3\text{V}_2(\text{PO}_4)_3/\text{Li}_{1.3}\text{Al}_{0.3}\text{Ti}_{1.7}(\text{PO}_4)_3$	100	$4.4 \times 3.0 \times 1.1$	40	—	-40 to $+80$ °C
Cymbet Corporation	EnerChip Bare Die CBC050 [327]	$\text{LiCoO}_2/\text{LiPON/Li}$	50	$5.7 \times 6.1 \times 0.2$	—	$>5\,000$ at 10% discharge	-40 to $+70$ °C
Front Edge Technology, Inc.	NanoEnergy [328]	$\text{LiCoO}_2/\text{LiPON/Li}$	100–1,000	$20 \times 25 \times (0.1-0.3)$	—	$<10\%$ loss after 1 000 cycles	-40 to $+100$ °C
Infinite Power Solutions	Thinergy MEC225 [329]	$\text{LiCoO}_2/\text{LiPON/Li}$	130	$12.7 \times 12.7 \times 0.17$	125	10 000	-40 to $+85$ °C
STMicroelectronics	EnFilm EFL700A39 [330]	$\text{LiCoO}_2/\text{LiPON/Li}$	700	$25.7 \times 25.7 \times 0.22$	200	$<20\%$ loss after 4,000 cycles	-40 to $+60$ °C

and Li metal as anode had achieved a cycle life of 10 000 with 90% capacity retention [305]. Anode-free configuration can be seen as a special type of Li metal anode. As the name suggests, anode-free batteries do not have an anode when it is been manufactured. Its Li metal anode forms during the charging process. This could lower the manufacturing cost with improved safety [321]. An uneven lithium deposition (dendrite) is the major challenge of anode-free batteries, because the contact between the SE and current collector may be insufficient [322]. The large volume change of Si also plagues Li anode/anode-free TFLBs, as the entire active layer is consumed and redeposited during a discharge/charge cycle.

3.1.2. Devices

Although the first secondary TFLB was developed more than three decades ago [323], only a few TFLBs are commercially available. Table 8 lists some commercially available TFLBs. The capacities of these commercial TFLBs are highly flexible, varying from tens to thousands μAh , which corresponds to the power consumption of lower-power IoT devices with a typical energy demand of several microwatts (μW s) to a few milliwatts (mW)s. The composition of these TFLBs is mostly $\text{LiCoO}_2/\text{LiPON/Li}$, which attributes to the high energy density of Li metal/ LiCoO_2 and the chemical/mechanical stability of LiPON. Despite its low ionic conductivity, LiPON is still the mainstream of TFLBs, showing the lack of development of other SE materials (NASICON, perovskite, Garnet). Among these TFLBs, only the EnerChip Bare Die batteries from Cymbet Corporation are designed for implementation in embedded systems (can be placed directly on a PCB), which is manufactured on p-type silicon wafers. Others are implemented as surface-mounted devices in IoT devices. One example of the powering IoT devices using thin film batteries is radio-frequency identification (RFID) tags, which are used in shipping and inventory control. Some of the RFID tags include sensors that can act as IoT nodes, and these tags are so thin that regular methods are not applicable in such case [324].

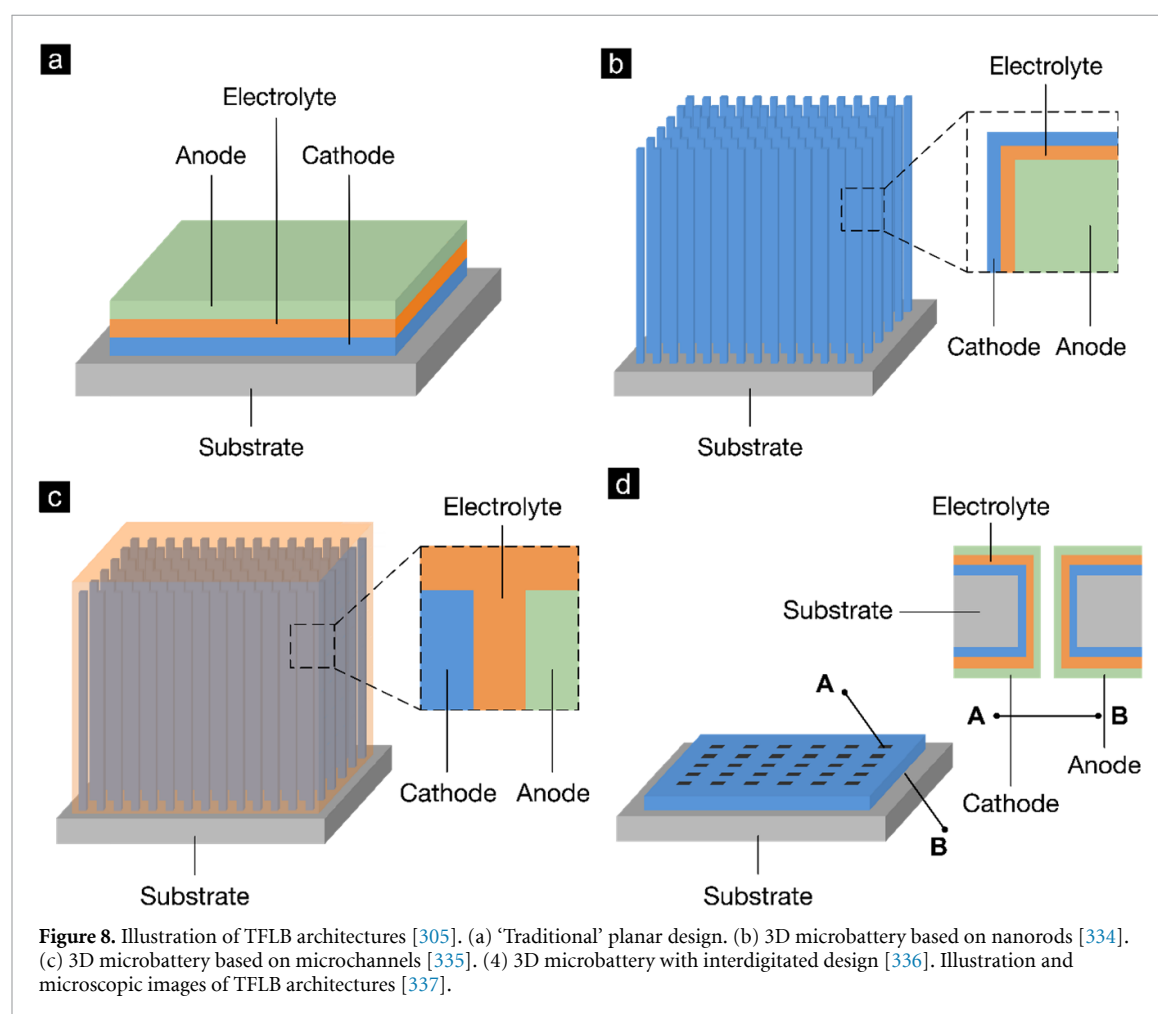
Some examples of reported TFLBs are also discussed here, as shown in table 9. Again, the electrolyte material for those TFLBs is LiPON, and for anode material is Li metal. A recent study utilizing LLZO electrolyte has shown an increased areal and gravimetric capacity compared to LiPON electrolyte [325]. The capacity retention of a polycrystalline LLZO electrolyte TFLB is still far away from that of a TFLB with an amorphous LiPON electrolyte.

3.1.3. Outlook

Although there are already some TFLBs commercially available, their specific energy and energy density is orders of magnitude lower than conventional pouch/cylindrical cells. The energy density of the TFLBs listed in table 8 ranges from 3.7 to 30 Wh kg^{-1} (10–86 Wh l^{-1}). In comparison, an LG INR 18650 MJ1 cylindrical cell (3500 mAh) has an energy density of 257 Wh kg^{-1} (710 Wh l^{-1}) [338]. Micro-batteries have lower energy density due to their lower active material/inactive material ratio. Nonetheless, the energy density of TFLBs still needs to be increased to improve the overall performance of micro- harvest/storage devices. In addition, the areal power capability of TFLBs is another limiting factor of their application to IoT devices. One method of increasing the areal power capability is replacing LiPON electrolyte with other SEs with a higher ionic conductivity, such as LLZO. This will need to solve the interfacial stability issue which could be challenging for both TFLBs and regular-sized batteries. A recent study realized an ultrathin amorphous

Table 9. Examples of reported TFLBs.

Year	Composition	Gravimetric capacity (mAh g^{-1})	Areal capacity ($\mu\text{Ah cm}^{-2}$)	Thickness (μm)	Voltage (V)	Cycle life	Deposition	References
2000	LCO/LiPON/Li	137	—	15	3.0–4.2	<2% loss after 4,000 cycles	RF sputtering	[331]
2014	LNMO/LiPON/Li	122	—	3	3.5–5.1	10% loss after 10 000 cycles	RF sputtering	[305]
2018	LTO/LiPON/Li	162	2.7	—	1.0–2.0	<10% loss after 200 cycles	Flame aerosol	[332]
2019	LCO/LiPON/ Anode-free	110	5.7	ca. 9.7	2.0–4.2	<20% loss after 100 cycles	RF sputtering & aerosol	[333]
2020	LCO/Li-Nb-O/LLZO half-cell	142	20.9	0.5	3.0–4.25	>80% after 138 cycles	RF sputtering	[325]



LLZO electrolyte, which has shown its ability to block the growth of Li dendrite like LiPON thin films [295]. Another approach to improving the capacity and rate capability is to fabricate TFLBs with 3D architectures. Xia *et al* reported self-standing 3D cathodes for TFLBs with improved interface kinetics. They deposited 3D LiMn_2O_4 (LMO) nanowall arrays and planar LMO thin films on conductive substrates using magnetron sputtering. The 3D TFLB has shown a superior capacity and rate capability compared to the 2D TFLB with discharge capacities of 119 and 83 mAh g^{-1} at 2 and 20 C rates, in contrast to 87 and 16 mAh g^{-1} of the 2D TFLB. Figure 8 illustrated some TFLB designs with different 3D architectures.

3.2. Micro-fuel cells for on-demand powering of IoT devices

Batteries represent by far the most common energy storage and powering concept for IoT devices. However, recent developments in micro-fabrication of electrochemical cells, such as fuel cells and electrolyzers, enable

a new perspective on the integration of fuel cell technologies as powering device also in self-controlling or remote applications, providing a compensation of power limits and intermittency of direct energy harvesting. By definition, a fuel cell is neither an energy storage nor an energy harvesting device. However, when integrated with a fuel reservoir/storage they hold the potential to work as an on-demand powering device like a battery, and at the same time, may harvest excess energy to internally recover their fuel via electrolysis (in reverse mode). Microscaled fuel cells and electrolyzers may hence complement batteries as a sustainable on-demand powering device based on hydrogen technology. However, such device concepts typically rely on an active supply of a chemical fuel, limiting their operation to niche applications. For example, IoT applications that allow external access for recharging/refueling, but require fast charging times (e.g. a hearing aid as medical device [339]), can benefit from micro fuel cell technologies. Hydrogen-based fuel cells were also proposed as a competitive powering solution for different types of IoT systems [340] and also explored in current European initiatives such as Harvestore [10, 341, 342].

Fuel cell devices are based on the idea of converting chemical energy into electrical output by forcing complementary redox half-reactions and subsequent ion transport across a membrane. Electrochemical energy conversion typically reflects a reaction of hydrogen and oxygen in the overall fuel cell reaction, leaving behind water as the main reaction product. As a result, the ion transport required for fuel cell operation is either proton transport or oxygen ion transport. Typical fuels are gaseous hydrogen or methane. The basic operation of fuel cells is CO₂-emission-free, making its technology sustainable and in principle green. Miniaturization of fuel cells and particularly micro-scale solid oxide fuel cells may be considered a key ingredient for their potential application in remote IoT devices. The ability to structure and build fuel cells based on nano- to microscale thin films allows the compact design ($\sim 1\text{ cm}^3$) of fuel cell devices that are compatible with the need of IoT devices.

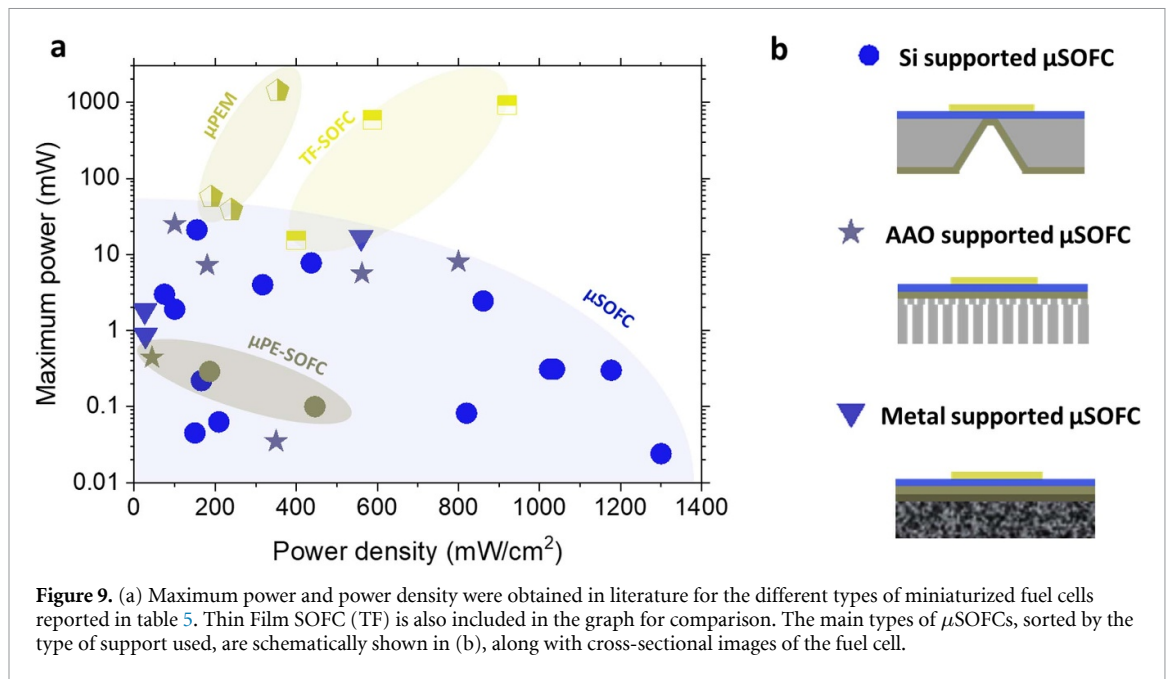
3.2.1. Materials

In general, the two main types of micro fuel cells for portable application are micro proton-exchange membrane fuel cells (μ PEM) and micro Solid oxide fuel cells (μ SOFC). In μ PEMs, the electrolyte is commonly a proton-conducting polymeric membrane that allows the exchange of ions between the oxidizing atmosphere at the cathode electrode and the gaseous fuel flowing through the anode electrode [343]. The high protonic conductivity of polymeric membranes such as Nafion® allows PEM to operate between ambient temperature and 90 °C, while their high humidity content prohibits the use above the water boiling point. This operational range permits a facile integration of μ PEM into standard Si microfabrication techniques, as was demonstrated by different authors [344–347].

Another promising type of miniaturized power generator for portable application is micro Solid oxide fuel cells (μ SOFCs) [348, 349]. μ SOFCs are characterized by the use of an oxygen ion conducting metal oxide thin film as an electrolyte. While in traditional SOFC temperatures higher than 700 °C are needed to overcome the ohmic losses derived by the poor ionic transport, nanometric electrolyte thin films can drastically reduce the operating temperature of μ SOFCs down to 400 °C, making them accessible to remote applications where waste heat is available in large amounts [350]. Research and development efforts toward nanoscale manipulation of ionic properties of materials principally allow realizing micro-scale membranes and exchange electrodes, building a fully operational fuel cell on the micrometer lengths scale. In addition to that, a vast variety of research is devoted to harvesting nanoscale materials science for improved fuel cell performance, including tailored nano-composite materials [351–353] and heterogeneously layered superlattices [354–358], or interface concepts with tailored ion conduction along and across interfaces [357, 359–361]. The interested reader is referred to Garbayo *et al* [362], Wen *et al* [363] and Shin *et al* [364] for more details on the strategies for improving the performances of electrode and electrolyte thin films for μ SOFC.

3.2.2. Devices

Table 10 shows a collection of characteristics, operating conditions, and performances reported in the literature for different types of μ SOFCs and μ PEMs. For comparison, table 10 is also reported a selection of results obtained for proton-conducting μ SOFCs (PE- μ SOFCs), in which the electrolyte is based on a protonic conducting oxide thin film, and thin films based SOFC (TF-SOFC), where thin films components are integrated with a bulk anode supported fuel cells. To better visualize the large spreading of μ SOFC performances, figure 9 shows the specific power density and total power reported in table 10. In the figure 9(b), the performances of μ SOFCs are also divided into the three main substrates used, i.e. (a) Microfabricated Si substrates, (b) Porous Anodized Aluminum oxides (AAO) and (c) Porous Metallic substrates. μ SOFCs reach very high values of power densities that place them among the most efficient micropower systems [348, 349]. Yet, the increase in power density is also observed to go along with a decrease



in total power measured, meaning that such power densities are mainly achievable when using very small active areas. This limitation is not due to poor performances of the μ SOFC components, since TF-SOFC can deliver power up to Watts, rather to the integration of dense thin films into porous substrates or self-sustained membranes, which may result in gas leakages or short circuits when large active areas are used. Another challenge that remains to be fully addressed is the integration of the fuel cell with the other auxiliary components, such as fuel micro-reformer or post combustor unit, in a compact and miniaturized system [365, 366].

3.2.3. Outlook

For rechargeable on-demand powering devices in IoT applications, fuel cell concepts may be preferred against battery solutions, when fast charging times are required or their implementation is beneficial for specific device functionality. Therefore, fuel cells have found their way into commercialization in small devices, with a prime example of a methanol fuel-cell-powered hearing aid brought into the market by Widex Inc.[®] (Widex Energy Cell[®]), which allows recharging of the methane-based PEM power supply within 20 s, rather than within hours for typical battery-power devices, and allowing 24 h of operation [367].

Finally, we notice that one of the main appealing features of micro fuel cells is the large energy density deriving from the use of liquid or gaseous fuels (up to 1 Wh L^{-1}), which is expected to allow a long off-grid (off-refueling) operation [368]. Yet, the employment of fuels may also represent a limitation when powering remote IoT devices, since the refilling of the fuel tank would be needed to 'recharge' the μ SOFCs. As a possible solution, a micro Solid Oxide Electrolysis Cell may be envisioned for promoting the reverse electrochemical reaction and generating a fuel, such as a hydrogen from environmental humidity, when excess energy from neighboring energy harvesting devices is available. In this way, a long-term operation in remote IoT devices may be achieved. Such remote electrolyzer technology, however, is yet in the early stage of research. Nevertheless, electrolyzer functionality is in principle accessible by running micro fuel cells in 'reverse' mode, to switch electrochemical conversion from fuel to electrical power in opposite direction and thus generating fuel from water. A futuristic prospect for a remote IoT device may be equipped with such a bifunctional electrochemical cell, capable of burning fuel when electrical power is required and producing fuel when excess electrical energy is available. Systemically collecting water by a capillary or microfluidic collector device may e.g. allow providing humidity in a small and controlled amount to the electrolyzer to restore the fuel level of the devices.

At this stage, the application of fuel cells and electrolyzers may be considered a niche in current IoT technologies. However, increasing degrees of miniaturization further enhanced power density, and increased conversion efficiency can be expected from fundamental research in the coming years, making such technologies a valuable member of the IoT toolbox.

Table 10. Comparison among performances, operating conditions, and characteristics of a selection of micro fuel cells presented in the literature. Notation: Si: Silicon supported, STS: Stainless steel, AAO: anodized aluminum oxide, YSZ: Ytria stabilized zirconia, LSCF: Sr doped Lanthanum cobalt ferrite, LSC: Strontium doped lanthanum cobaltite, GDC: Gadolinium doped Ceria, SDC: Samarium doped Ceria, BZY: Yttrium doped Barium Zirconate.

	References	Year	Substrate	Composition Cathode/ electrolyte/ anode	Op. T (°C)	Max. Power (mW)	Max. power density (mW/cm ²)	Active area (mm ²)	Current at power peak (A cm ⁻²)	Voltage at current peak (V)	V _{oc} (V)
μ SOFC	An <i>et al</i> [369]	2013	Si, corrugated	Pt/YSZ-YDC/Pt	450	0.024	1300	0.002	4.1	0.32	1.05
	Chao <i>et al</i> [370]	2011	Si, corrugated	Pt/YSZ/Pt	450	0.082	820	0.01	3	0.27	1.11
	Su <i>et al</i> [370]	2008	Si, corrugated	Pt/YSZ/Pt	450	2.44	861	0.3	2.7	0.318	1.09
	Tölke <i>et al</i> [371]	2012	Foturan®	Pt/YSZ/Pt	550	0.063	209	0.03	0.62	0.34	0.57
	Muecke <i>et al</i> [372]	2008	Foturan®	Pt/YSZ/Pt	550	0.045	150	0.03	0.36	0.42	0.77
	Schlupp <i>et al</i> [373]	2014	Si	Pt/YSZ/Pt	410	0.22	166	0.13	0.47	0.35	0.84
	Kerman <i>et al</i> [374]	2011	Si	Pt/YSZ/Pt	500	0.31	1037	0.03	3.3	0.314	0.97
	Tsuchiya <i>et al</i> [375]	2011	Si	Pt/YSZ/LSCF	510	21.1	155	25	0.46	0.34	0.75
	Kerman <i>et al</i> [376]	2012	Si	Pt/YSZ-CGO/Pt	510	0.31	1025	0.03	4.4	0.23	0.41
	Kerman <i>et al</i> [377]	2012	Si	Pt/CGO/Pt (Ru)	520	0.3	1177	0.03	4	0.29	0.6
	Kerman <i>et al</i> [378]	2015	Si	Pt/YSZ/Pt	500	3	75	3.14	0.25	0.3	0.87
	Garbayo <i>et al</i> [379]	2014	Si	LSC/YSZ/CGO	750	1.9	100	1.9	0.17	0.58	1.1
	Baek <i>et al</i> [380]	2015	Si	Pt/YSZ/Pt	400	7.72	437	17.6	1.3	0.34	1.1
	Baek <i>et al</i> [381]	2016	Si	Pt/YSZ/Pt	400	4	317	1.2	1.1	0.29	1.04
	Kim <i>et al</i> [382]	2016	STS/YSZ	LSC/YSZ/Ni-YSZ	550	16.8	560	3	1.3	0.46	1.1
	Joo <i>et al</i> [383]	2008	Ni	Pt-LSC/GDC/Ni	450	1.82	26	7	0.07	0.37	0.87
	Lee <i>et al</i> [384]	2014	Ni/STS	Pt-LSC/YSZ-GDC/Ni-GDC	450	0.87	28	3.1	0.07	0.4	0.91
	Kwon <i>et al</i> [385]	2011	AAO	Pt/YSZ/Pt	500	0.035	350	0.01	1.2	0.29	1.02
	Park <i>et al</i> [386]	2014	AAO	Pt/YSZ-GDC-YSZ/Pt	500	25	100	25	0.25	0.40	1.1
	Oh <i>et al</i> [387]	2018	AAO	Pt/SDC-YSZ-SDC/Pt	450	5.62	562	1	1.25	0.45	1.07
	Ha <i>et al</i> [388]	2013	AAO	Pt/YSZ/Pt	450	7.2	180	4	0.37	0.48	1.14
	Shin <i>et al</i> [389]	2019	AAO	Pt/SDC-YSZ-SDC/Pt-Ceria	500	8	800	1	1.8	0.44	1.14
μ -PC-SOFC	Li <i>et al</i> [390]	2016	Si	Pt/GDC-BZY/Pt	425	0.1	446	0.022	2.25	0.20	1.1
	Kim <i>et al</i> [391]	2011	Si, corrugated	Pt/BZY/Pt	450	0.29	186	0.16	0.9	0.21	1.0
	Chang <i>et al</i> [392]	2013	AAO	Pt/BZY/Pt	450	0.44	44	1	0.12	0.37	1.0
TF-SOFC	Reolon <i>et al</i> [393]	2018	NiO-YSZ	Pt-LSC/CGO-YSZ/Ni-YSZ	650	15.4	400	38.5	0.7	0.57	0.91
	Shin <i>et al</i> [384]	2020	NiO-YSZ	LSC/CGO/Ni-YSZ	500	920	920	100	1.7	0.54	1.14
	Noh <i>et al</i> [394]	2014	NiO-YSZ	LSC-GDC/CGO-YSZ/Ni-YSZ	500	588	588	100	1.2	0.49	1.1
μ -PEM	Yu <i>et al</i> [347]	2019	Si	Pt-C/Nafion/Pt-C	80	1416	354	400	0.8	0.44	0.98
	Yu <i>et al</i> [344]	2003	Si	Pt-C/Nafion/Pt-C	25	57	190	30	0.26	0.73 (2 cells)	1.9
	Morisawa <i>et al</i> [347]	2014	Si	Pt-C/Nafion/Pt-C	45	38.4	240	16	0.4	0.5	1

4. Section-III: power management for IoT applications

Power management units (PMUs) are rapidly being developed to provide a significant amount of power for sensors in IoT networks. The DC–DC power converter is the main component in the PMU placed in between the power sources and the IoT devices. The aims of the DC–DC converters are to:

- **Regulate the input voltage coming from the sources to a required output voltage level to operate the IoT device;** Match the source impedance with the IoT device impedance to achieve the maximum power transfer.

To maintain a long battery or sustainable power sources life, high power-conversion efficiency is critical for the PMU since it can add substantial overhead to the overall power consumption of the chip.

4.1. DC–DC power converters

IoT applications operate at low supply voltages which often range below 1 V, while the battery as the primary or auxiliary power source for their stable voltage supply provides a voltage range from 1.2 V to 5 V. This requires DC–DC converters to step down voltages. Low dropout (LDO) regulators [395, 396] provide high supply rejection and regulated output voltage. However, this approach results in poor power efficiency. Buck converters are a common architecture in step-down voltage converters for higher power regulations [397, 398]. Because of the large overhead associated with its control [398] along with the inductor size, buck converters are also not good candidates for low power applications. Another approach is to use a switched-capacitor DC–DC converter which has a power efficiency of up to 97% [399]. However, switched capacitor designs typically only divide the input voltage and do not provide the ability of voltage regulation [400]. A hybrid architecture has been shown in [401–403] where the first stage divides the battery voltage by five using a switched capacitor DC–DC converter, then the design uses an LDO in the second stage for output voltage regulation. However, the two-stage approaches suffer a lower conversion efficiency. It has been shown to achieve maximum efficiency, where the hybrid architecture regulates a 1.3 V input to 0.4 V and achieves a power efficiency of >80% for load ranges 1–240 μ A [401].

4.2. Maximum power point tracking (MPPT) control

When electrically connected to an external load, the power energy sources such as solar and TEGs have power characteristic curves where the impedance of the external load has a particular unique size, maximum power is retrieved. This unique impedance varies considerably with both the design of the power energy sources and their operating conditions. Subsequently, a MPPT control method needs to be utilized to actively emulate the impedance of DC–DC to match the inner impedance of the power energy sources so that the sources are always operating at the maximum power point.

State-of-the-art solutions are capable of performing MPPT and/or charging capabilities with dc renewable sources [404–408]. As the I – V and P – V characteristics of the TEG follow the same principles as that of the solar panel, MPPT techniques that are commonly used for solar panels can be reused for the TEG, for instance, the perturb and observe (P&O) algorithm, hill climbing, and incremental conductance technique. One of the most popular MPP finding methods in IoT applications is called fractional open-circuit voltage (FOCV)[408]. FOCV method samples the open-circuit voltage (VOC) of the energy harvester first and then determines the ratio of the VOC, which corresponds to the maximum power point of the energy harvester. It is an *a priori* technique based on many past analyses and experiments with the findings that the MPP for PV cells is around 0.71–0.82 of their VOC while other types of energy sources, such as RF rectennas, TEGs, PEHs, and EMGs, are around half of the open-circuit voltage (VOC/2). The PMU proposed in [406] is comprised of a $10\times$ step-up charge pump with two-dimensional (2-D) tuning for MPPT and a digital LDO regulator with input power sense capabilities.

4.3. Challenges and trends

- The single battery source-based IoT system falls into disuse. The more renewable energy sources with a combination of battery sources start to play a mainstreaming to maintain long battery or sustainable power sources' life. This requires new circuit architectures with multiple inputs for power management to efficiently deal with the power.
- Providing both MPPT as well as output voltage regulation is still an issue to be fully resolved. The state of the art still used two-stage power converters. The first stage is usually buck, boost, or switched capacitor converters with the implementation of MPPT power transfer, and the second stage is usually the LDO or another DC–DC converter to regulate the output voltage fitting within the applications. This two-stage solution gives a poor PCE and a relatively larger size of the PMU. There might be a chance for researchers to integrate as a single-stage circuit architecture for both MPPT and output voltage regulation functions.
- The PMU requires low power consumption, small chips, small external components, such as inductors and capacitors, and high efficiency. Therefore, low power and high-efficiency power management integrated circuit (PMIC) is important to extend battery life and sustain the system without charge. The main challenge is how to integrate bulk components such as inductors and capacitors in the same silicon substrate. A novel three-dimensional in-silicon through-silicon via (TSV) magnetic-core toroidal inductor for power supply

in packing is proposed in [409], and the first interposer converter with 3D TSV inductor is demonstrated at DTU [410].

5. Concluding remarks

Harvesting and storage of energy play an important role in the efficiency and lifetime of IoT devices. The central goal of energy harvesting systems for IoT is to move from battery-based devices to an autonomous energy harvesting system that relies on energy harvested from the ambient environment. To reduce the power consumption of batteries, a crucial step is to find alternative harvesting and storage solutions and optimize the PMICs, which helps to enhance the system's life span. In this review paper, we summarized the current state-of-the-art IoT enabling and emerging harvesting and storage technologies covering the areas of energy harvesting- and storage devices, and power management for IoT applications.

To harvest energy from a sustainable power supply, it is important to ensure the availability of the energy source from which energy is supposed to be harvested, the high amount of harvested energy, and the high efficiency of the harvesting system just to name a few. In this review, we have provided some of the limitations and solutions concerning the energy sources and the way to harvest these. However, in reality, the sources of energy are not continuous, and hence, a new way to ensure the continuous operation of IoT devices should be considered. Recent advancements in IoT have drawn the attention of researchers and developers toward systems where the storage unit is installed together with the harvesting units in the same device [9, 281, 282, 411]. This surely increases the complexity of the IoT devices from the viewpoint of electronics, when compared to the electronics associated with only battery-powered IoT devices, but on the other hand, it provides an efficient and autonomous solution.

Furthermore, most of the harvesting devices discussed in this paper show the technological potential to replace batteries for powering the IoT from the viewpoint of power output per unit device cross-section area. The integration of high density of harvesting devices with a small footprint area is required to reach the actual requirements of the power needed for the devices during the different modes of their operation. Finally, the cost factors, such as the cost of the harvester system as compared with the battery storage must also be considered when designing such a system [29]. Despite having still a vast amount of challenges in the integration of the energy harvesting devices into the IoT units and their operation, they remain one of the most reliable practical solutions for the replacement of batteries.

Data availability statement

The data that support the findings of this study are available upon reasonable request from the authors.

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Author contributions

Initiating the idea, structuring the review article, abstract, introduction, and concluding remarks- A C, and N P; Thermoelectricity for IoT- A C, C N L, and N P; Piezoelectricity for IoT- H Z, A B and V E; Flexoelectricity for IoT- S Y, A R I, and D V C; Electromagnetic energy harvester- C I and R B; Photovoltaics for IoT- H A, M A, C Y H, and M M; Battery for IoT- J C and P N; Fuel cells and electrolyzers for IoT- F M C and F G; Power management for IoT applications- Z O. All authors reviewed and edited the manuscript.

Conflict of interest

Authors declare no conflict of interest.

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